

# UNIVERSITY OF TRENTO

#### DIPARTIMENTO DI INGEGNERIA E SCIENZA DELL'INFORMAZIONE

38123 Povo – Trento (Italy), Via Sommarive 14 http://www.disi.unitn.it

#### ADS-BASED GUIDELINES FOR THINNED PLANAR ARRAYS

G. Oliveri, L. Manica, and A. Massa

January 2011

Technical Report <u># DISI-11-091</u>

# **ADS-Based Guidelines for Thinned Planar Arrays**

G. Oliveri, Student Member, IEEE, L. Manica, and A. Massa, Member, IEEE

ELEDIA Research Group Department of Information and Communication Technologies, University of Trento, Via Sommarive 14, 38050 Trento - Italy Tel. +39 0461 882057, Fax +39 0461 882093 E-mail: andrea.massa@ing.unitn.it, {giacomo.oliveri, luca.manica}@dit.unitn.it Web-site: http://www.eledia.ing.unitn.it

### **ADS-Based Guidelines for Thinned Planar Arrays**

G. Oliveri, L. Manica, and A. Massa

#### Abstract

This paper proposes an analytical technique based on Almost Difference Sets (ADSs) for thinning planar arrays with well controlled sidelobes. The method allows one to synthesize bidimensional arrangements with peak sidelobe levels (PSLs) predictable and deducible from the knowledge of the array aperture, the filling factor, and the autocorrelation function of the ADS at hand. The numerical validation, concerned with both small and very large apertures, points out that the expected PSL values are significantly below those of random arrays and comparable with those from different sets (DSs) although obtainable in a wider range of configurations.

**Key words**: Array Antennas, Planar Arrays, Thinned Arrays, Sidelobe Level Control, Almost Difference Sets.

### **1** Introduction

Antenna arrays for radar tracking, remote sensing, biomedical imaging, satellite and ground communications have often to support three-dimensional scanning with a suitable beampattern shape in the whole angular region [1]. Towards this end, planar arrays have to be used and large apertures are necessary to provide satisfactory angular resolutions along both azimuth and elevation [1]. On the other hand, the inter-element spacing should not exceed half-wavelength to avoid the presence of grating lobes [1]. These requirements usually result in very inefficient, heavy, and expensive solutions consisting of planar geometries with several thousands close elements.

In order to reduce the number of elements while keeping the radiation properties of the original structures, thinning techniques have been successfully introduced [2]. Designing thinned planar arrays is an important research topic since decades (see [2][3][4][5][6][7][8] and the references cited therein). As a matter of fact, a suitable thinning allows one to reduce the array costs, its weight, and the power consumption. However, it causes the loss of the control of the peak sidelobe level (*PSL*) [6] to be properly counteracted. To this end, several techniques has been proposed in order to fully exploit the advantages of thinned arrangements while minimizing their drawbacks. First attempts have been conceived to require low computational resources (see Tab. I in [9]), but they have provided no significant improvements when compared with random placements [2][9] extensively employed in practice [2].

More recently, the availability of large computational resources has justified the use of optimization techniques such as dynamic programming [10], genetic algorithms [5][11][8], simulated annealing [12][13], and particle swarm optimizers [7]. Thinned arrays synthesized with optimization tools turn out to be very effective [10][12][7][8], even though it is not possible to *a-priori* estimate the expected performances for a given array aperture and thinning factor [6]. Furthermore, computational and convergence issues make the application of stochastic optimizers difficult and expensive when dealing with 1D large apertures [6] and, even more, when planar arrangements are considered.

In order to overcome such drawbacks, an alternative approach for thinning large arrays has been introduced [4][6][14]. Such an approach relies on the exploitation of binary sequences

derived from Difference Sets (DSs), which exhibit a two-level autocorrelation function [4]. Besides their analytic nature and the arising inexpensive generation, DS-based thinned arrays have several interesting properties. They are deterministically designed and present predictable [6] and low PSLs (3 dB and 1.5 dB below random arrays for the linear case and the planar one, respectively). However, only a limited number of DS sequences exist and the whole set of aperture sizes and thinning values [6][15] cannot be dealt with.

The problem of obtaining sub-optimal sequences (in terms of autocorrelation levels) has been recently addressed in information theory and "close" sequences to DSs have been looked for. Almost Difference Sets (ADSs) [16][17][18] are a wide class of binary sequences with *three-valued* autocorrelations [16][17][18]. They represent the closest sets to DSs [16][17][18] (three-levels vs. two-levels) and large repositories of explicit sequences (e.g., [19]) are available.

As regards to 1D geometries, the sidelobe characteristics of ADS-based arrays have been analyzed in [20] and good performances have been predicted and numerically verified dealing with both small and large apertures. Because of these results and its deterministic nature, an ADSbased technique seems to be a good candidate for thinning planar arrangements of radiating elements and it will be presented in this paper. More specifically, the objective is not to define the "optimal" thinning method, but rather to provide a simple and reliable technique which guarantees to the designer predictable performances to be taken into account during the feasibility study. Towards this end, the PSL behavior of ADS-based planar arrays will be analytically investigated and different bounds will be provided. It should be pointed out that, despite the linear case [20] where the Blahut's theorem [20] has been applied, a different mathematical analysis is here necessary. The PSL bounds are then derived starting from the properties of the 2D discrete Fourier transform.

The paper is organized as follows. After a short overview on ADSs (Sect. 2), a set of suitable bounds of the PSL are analytically determined in Sect. 3. Section 4 provides a selected set of numerical results aimed at validating the obtained PSL estimators as well as comparing the ADS performances with both random techniques and state-of-the-art optimization approaches. The exploitation of directive elements is also considered in order to point out the flexibility of the ADS thinning theory. Finally, some conclusions are drawn (Sect. 5).

### 2 Two-Dimensional Almost Difference Sets

With reference to the 2D problem, let us define a  $(PQ, K, \Lambda, t)$ -almost difference set as a Ksubset  $\underline{\mathbf{D}} = \{ \mathbf{d}_k \in \underline{\mathbf{G}}, k = 0, ..., K - 1 \}$  of the Abelian group  $\underline{\mathbf{G}}$  of order PQ ( $\underline{\mathbf{G}} \triangleq \mathbb{Z}^P \otimes \mathbb{Z}^{Q(1)}$ , P and Q being chosen according the Kronecker Decomposition Theorem [21]) for which
the multiset

$$\underline{\mathbf{M}} = \{\mathbf{m}_{j} = (\mathbf{d}_{h} - \mathbf{d}_{\ell}), \mathbf{d}_{\ell} \neq \mathbf{d}_{h}; j = 0, ..., K(K - 1) - 1\}$$

contains t nonzero elements of <u>G</u> each exactly  $\Lambda$  times and the remaining PQ - 1 - t nonzero elements each exactly  $\Lambda + 1$  times [18]. Therefore, an *ADS* satisfies the following existence condition [17][18]:

$$K(K-1) = t\Lambda + (PQ - 1 - t)(\Lambda + 1) \tag{1}$$

where  $K \ge \Lambda + 1$ ,  $0 \le K \le PQ$ , and  $0 \le t \le PQ - 1$ . Moreover, it is worth noticing that DSs are ADSs for which t = PQ - 1 or t = 0 [18].

If  $\underline{\mathbf{D}}$  is a  $(PQ, K, \Lambda, t)$ -ADS, then it is possible to derive a two dimensional binary sequence  $\underline{\mathbf{W}} = \{w(p,q) = 1(0) \text{ if } (p,q) \in (\notin)\underline{\mathbf{D}}; p = 0, ..., P - 1, q = 0, ..., Q - 1\}$  whose 2D periodic autocorrelation function [6]  $A_w(p,q)$  ( $p \in [0, P - 1], q \in [0, Q - 1]$  being its periodicity) is a three-level function [16][18]

$$A_w^{ADS}(p,q) = (K - \Lambda)\delta(p,q) + \Lambda + \sum_{r=1}^{N-1-t} \delta(p - l_{r,1}, q - l_{r,2}),$$
(2)

where  $K \ge \Lambda + 1$ ,  $\delta(p,q) = 1$  if p = q = 0 and  $\delta(p,q) = 0$  otherwise, and  $(l_{r,1}, l_{r,2}) \triangleq \mathbf{l}_r$ is an element of the set  $\underline{\mathbf{L}} = \{\mathbf{l}_r \in \mathbb{Z}^P \otimes \mathbb{Z}^Q, r = 1, ..., N - 1 - t\}$ . For descriptive purposes, let us consider the ADSs in Tab. I [18][19]. The plots of  $\underline{\mathbf{W}}$  and of the three-level function  $A_w^{ADS}(p,q)$  in correspondence with  $\underline{\mathbf{D}}_i$  (i = 1, 2, 3) are shown in Fig. 1.

<sup>&</sup>lt;sup>(1)</sup> The symbol  $\otimes$  stands for the *direct sum* of  $\mathbb{Z}^P$  and  $\mathbb{Z}^Q$ , that is  $\underline{\mathbf{G}} = \{\mathbf{g}_i = (\alpha_j, \beta_h), \alpha_j \in \mathbb{Z}^P, \beta_h \in \mathbb{Z}^Q, i = 0, ..., PQ - 1, j = 0, ..., P - 1, h = 0, ..., Q - 1\}$  and  $\underline{\mathbf{G}}$  is equipped with the component-wise operations derived from  $\mathbb{Z}^P$  and  $\mathbb{Z}^Q$ , that is  $\mathbf{g}_1 + \mathbf{g}_2 = ((\alpha_1 + \alpha_2)_{mod P}, (\beta_1 + \beta_2)_{mod Q}).$ 

As regards to the closeness of the ADS to the DS sequences, likewise 1D arrangements, the bidimensional autocorrelation function of a  $(PQ, K, \Lambda, t)$ -ADS differs from  $A_w^{DS}(p, q) = K \delta(p, q) + \Lambda$  [6] by a unity in only PQ - 1 - t points [16][18] [Eq. (2)]. Moreover, the ADSautocorrelation function still remains unaltered after cyclic shifts of the reference sequence [16][18] since if **D** is an ADS, then

$$\underline{\mathbf{D}}^{(\sigma_x,\sigma_y)} = \{((p+\sigma_x)_{mod P}, (q+\sigma_y)_{mod Q}); (p,q) \in \underline{\mathbf{D}}, \sigma_x, \sigma_y \in \mathbb{Z}\}$$
(3)

is still an ADS. As a consequence, starting from a  $(PQ, K, \Lambda, t)$ -ADS, it is always possible to build  $P \times Q$  different  $(PQ, K, \Lambda, t)$ -ADS by applying cyclic shifts to its elements.

### **3** ADS-Based Planar Arrays - Mathematical Formulation

Let us consider a planar array of  $P \times Q$  elements located, according to the binary sequence  $\underline{W}$ , on a bidimensional lattice of points spaced by  $d_x$  and  $d_y$  wavelenghts along the x and y directions, respectively. The array factor of such an elements arrangement turns out to be [6][1]

$$AF\left\{\underline{\mathbf{W}}\right\} \triangleq W_{AF}(u,v) = \sum_{p=0}^{P-1} \sum_{q=0}^{Q-1} w(p,q) exp\left[2\pi j \left(pd_x u + qd_y v\right)\right]$$
(4)

where  $u = sin(\theta)cos(\phi)$  and  $v = sin(\theta)sin(\phi)$ . Moreover,  $W_{AF}(u, v)$  can be also expressed in terms of the 2D Discrete Time Fourier transform (DTFT) of the sequence  $\underline{W}$ ,

$$DTFT\left\{\underline{\mathbf{W}}\right\} \triangleq \mathbb{T}(\alpha,\beta) = \sum_{p=0}^{P-1} \sum_{q=0}^{Q-1} w(p,q) exp\left[-j\left(p\alpha + q\beta\right)\right],\tag{5}$$

as follows

$$W_{AF}(u,v) = \mathbb{T}(-2\pi d_x u, -2\pi d_y v).$$
(6)

Furthermore, by applying the Sampling Theorem [22] to the function  $\mathbb{T}(\alpha,\beta)$  ,

$$\mathbb{T}(\alpha,\beta) = \sum_{k=0}^{P-1} \sum_{l=0}^{Q-1} \mathbb{F}(k,l) \frac{\sin\left(\frac{\alpha P}{2} - k\pi\right)}{P\sin\left(\frac{\alpha}{2} - \frac{k\pi}{P}\right)} \frac{\sin\left(\frac{\beta Q}{2} - l\pi\right)}{Q\sin\left(\frac{\beta}{2} - \frac{l\pi}{Q}\right)},\tag{7}$$

 $\mathbb{F} \text{ being } 2D \text{ Discrete Fourier Transform } (DFT) \text{ of the sequence } \underline{\mathbf{W}} (DFT \{ \underline{\mathbf{W}} \} \triangleq \mathbb{F}(k, l) = \sum_{p=0}^{P-1} \sum_{q=0}^{Q-1} w(p, q) exp \left[ -2\pi j \left( \frac{pk}{P} + \frac{ql}{Q} \right) \right] \text{, it results that}$ 

$$W_{AF}(u,v) = \sum_{k=0}^{P-1} \sum_{l=0}^{Q-1} \mathbb{F}(k,l) \frac{\sin(\pi d_x u P + k\pi)}{P\sin(\pi d_x u + \frac{k\pi}{P})} \frac{\sin(\pi d_y v Q + l\pi)}{Q\sin(\pi d_y v + \frac{l\pi}{Q})}.$$
(8)

Such a relationship states that the samples of the array factor at  $u = \frac{k}{d_x P}$ ,  $v = \frac{l}{d_y Q}$  are equal to the values of the DFT of the weighting sequence  $\underline{\mathbf{W}}$  in (k, l)

$$W_{AF}\left(-\frac{k}{d_x P}, -\frac{l}{d_y Q}\right) = \mathbb{F}(k, l).$$
(9)

For illustrative purposes, Figure 2 shows the plot of the array factor  $(d_x = d_y = \frac{1}{2})$  and the samples of the *DFT* of <u>W</u> in correspondence with the set <u>D</u><sub>1</sub> [Fig. 2(*a*)] and the set <u>D</u><sub>2</sub> [Fig. 2(*b*)].

As regards to the peak sidelobe level (PSL), it is defined as

$$PSL\left\{\underline{\mathbf{D}}^{(\sigma_x,\sigma_y)}\right\} \triangleq \frac{max_{(u,v)\notin R} |W_{AF}^{(\sigma_x,\sigma_y)}(u,v)|^2}{|W_{AF}^{(\sigma_x,\sigma_y)}(0,0)|^2}$$
(10)

where  $W_{AF}^{(\sigma_x,\sigma_y)}(u,v)$  is the array factor coming from the shifted set  $\underline{\mathbf{D}}^{(\sigma_x,\sigma_y)}$  and R is the mainlobe region of extension (see Appendix)

$$R \triangleq \left\{ (u,v) \in [-1,1] \times [-1,1] : u^2 + v^2 \le 1, \, uv \le \frac{K}{4PQd_xd_y \max_{(k,l) \in \mathcal{H}_0} |\mathbb{F}(k,l)|} \right\}$$
(11)

with  $\mathcal{H}_0 \triangleq \underline{\mathbf{G}} \setminus (0,0)^{(2)}$ 

By substituting (8) in (10), it appears that

$$PSL\left\{\underline{\mathbf{D}}^{(\sigma_{x},\sigma_{y})}\right\} = \frac{\max_{(u,v)\notin R} \left|\sum_{k=0}^{P-1}\sum_{l=0}^{Q-1} \mathbb{F}^{(\sigma_{x},\sigma_{y})}(k,l) \frac{\sin(\pi d_{x}uP+k\pi)}{P\sin(\pi d_{x}u+\frac{k\pi}{P})} \frac{\sin(\pi d_{y}vQ+l\pi)}{Q\sin(\pi d_{y}v+\frac{l\pi}{Q})}\right|^{2}} = \frac{1}{K^{2}}\max_{(u,v)\notin R} \left|\sum_{k=0}^{P-1}\sum_{l=0}^{Q-1} \mathbb{F}^{(\sigma_{x},\sigma_{y})}(k,l) \frac{\sin(k\pi)}{P\sin(\frac{k\pi}{P})} \frac{\sin(\pi d_{x}uP+k\pi)}{Q\sin(\pi d_{x}u+\frac{k\pi}{P})} \frac{\sin(\pi d_{y}vQ+l\pi)}{Q\sin(\pi d_{y}v+\frac{l\pi}{Q})}\right|^{2}}$$

$$(12)$$

<sup>(2)</sup> The notation  $\underline{\mathbf{G}} \setminus (0,0)$  indicates the set of elements of the Abelian group  $\underline{\mathbf{G}}$  without the null element, (0,0).

since  $\mathbb{F}^{(\sigma_x,\sigma_y)}(0,0) = \sum_{p=0}^{P-1} \sum_{q=0}^{Q-1} w^{(\sigma_x,\sigma_y)}(p,q) = K$ ,  $\underline{\mathbf{W}}^{(\sigma_x,\sigma_y)} = \{w^{(\sigma_x,\sigma_y)}(p,q); p = 0, ..., P-1; q = 0, ..., Q-1\}$  being the two-dimensional sequence derived from  $\underline{\mathbf{D}}^{(\sigma_x,\sigma_y)}$ . As it can be noticed, the *PSL* of an *ADS*-based array is a function of the coefficients  $\mathbb{F}^{(\sigma_x,\sigma_y)}(k,l)$ . Unfortunately, since these coefficients cannot be expressed in closed-form (but their values are available when the generating *ADS* is known) and, unlike *DS*s, depends on the indexes k and l, it is not possible to provide a *PSL* threshold as for *DS*s-based planar arrays [6]. Nevertheless, the following set of inequalities holds true for sufficiently large values of P and Q (Appendix)

$$PSL_{INF} \le PSL_{min} \le PSL^{opt} \{\underline{\mathbf{D}}\} \le PSL_{max} \le PSL_{SUP}$$
(13)

where 
$$PSL^{opt} \{\underline{\mathbf{D}}\} = min_{(\sigma_x,\sigma_y)} \left[ PSL \left\{ \underline{\mathbf{D}}^{(\sigma_x,\sigma_y)} \right\} \right], PSL_{min} = \frac{\Omega\{\underline{\mathbf{D}}\}}{K^2}, PSL_{max} = \frac{\Omega\{\underline{\mathbf{D}}\}E\{\Gamma_{PQ}^{opt}\}}{K^2},$$
  
 $PSL_{INF} = \frac{K-\Lambda-\sqrt{\frac{(t+1)(PQ-1-t)}{PQ-1}}}{K^2}, PSL_{SUP} = \frac{(K-\Lambda+\sqrt{(t+1)(PQ-1-t)})E\{\Gamma_{PQ}^{opt}\}}{K^2}, \quad \Omega\{\underline{\mathbf{D}}\} = max_{(k,l)\in\mathcal{H}_0} \left| \mathbb{F}^{(\sigma_x,\sigma_y)}(k,l) \right|^2, \text{ and } E\{\Gamma_{PQ}^{opt}\} \approx -0.1 + 1.5 \log_{10}(PQ).$ 

It is now worth pointing out that  $PSL_{min}$  and  $PSL_{max}$  can be evaluated only once the ADS sequence is exactly known, since the knowledge of the term  $|\mathbb{F}^{(\sigma_x,\sigma_y)}(k,l)|^2$  is required, while the bounds  $PSL_{INF}$  and  $PSL_{SUP}$  can be *a-priori* determined starting for the knowledge of the characteristic parameters describing the ADS (i.e.,  $P, Q, K, \Lambda$ , and t).

For a preliminary validation of such an estimate criterion, let us refer to the planar array generated by  $\underline{D}_3$  in Tab. I. As expected, the PSL of the set  $\underline{D}_3^{(\sigma_x,\sigma_y)}$  depends on the values of  $\sigma_x$ and  $\sigma_y$  [Fig. 3(*a*)] and different shift values give the same optimal PSL,  $PSL_{opt}$ , whose value lies into the range of confidence defined in (13) [Fig. 3(*b*)]. The multiplicity of the optimal solutions indicates that less than  $P \times Q$  evaluations are actually needed to identify the optimal ADS-based planar array. This is a negligible computational cost compared to the burden required by stochastic optimization techniques to determine a thinned arrangement on the same aperture.

#### 4 Numerical Analysis

In this section, the results of a numerical assessment are described and discussed to point out potentialities and limitations of the ADS-based approach proposed as a suitable tool for predicting the performance of an effective set of planar thinned arrays. For comparison purposes, random arrangements [3][6] are considered as reference since, likewise ADS arrays, their performances can be *a-priori* estimated. More in detail, the estimator of the normalized peak sidelobe level of *planar random arrays* (RND) turns out to be [3]

$$PSL_{RND} = \frac{-ln\left[1 - \beta^{\frac{\lambda^2}{\pi^2 x_0 y_0(P-1)(Q-1)}}\right] + 1 - 2\left\{ln\left[1 - \beta^{\frac{\lambda^2}{\pi^2 x_0 y_0(P-1)(Q-1)}}\right]\right\}^{-1}}{K}$$
(14)

where  $\beta$  is the probability or confidence level that no sidelobe exceeds the  $PSL_{RND}$  value. Moreover, *random lattice planar arrays* (*RNL*), whose elements are located on a uniformlyspaced lattice of points over the aperture, exhibit the following *PSL* [6]

$$PSL_{RNL} = PSL_{RND} \times (1 - \nu) \tag{15}$$

(where  $\nu = \frac{K}{PQ}$  is the thinning factor).

The first numerical example deals with the analysis of the PSL bounds (13) versus  $\eta \triangleq \frac{t}{PQ-1}$ for different apertures and when  $\nu = 0.5$  (Fig. 4). As expected (Sect. 2), the upper bound of PSL tends to  $PSL_{DS}$  when  $\eta = 1$  and  $\eta = 0$  ( $PQ \to \infty$ ) and its value,  $PSL_{SUP}$ , is always below  $PSL_{RND}$  and  $PSL_{RNL}$  except for a small set of  $\eta$  values close to  $\eta = 0.5$  and large apertures ( $PQ \ge 10^6$ ). As a matter of fact, whatever the array dimension, the worst performances verify in correspondence with  $\eta = 0.5$ . Therefore, such an index value will be analyzed in the following to provide "worst-case" indications on ADS-based thinning.

Figure 5(c) shows the behaviors of the ADS bounds versus the aperture dimension ( $\nu = \eta = 0.5$ ). Since ADS are here available [19],  $PSL_{opt}$ ,  $PSL_{min}$ , and  $PSL_{max}$  are reported, as well. As it can be noticed, these plots confirm that  $PSL_{SUP}$  usually overestimates the actual peak sidelobe of the ADS array (while  $PSL_{min} \rightarrow PSL_{INF}$ ) and that  $PSL_{opt}$  is always well below the values exhibited by random families.

For completeness, the remaining of Fig. 5 gives an indication on the estimated behavior of ADS arrays in correspondence with different thinning percentages [ $\nu = 0.3$  - Fig. 5(*a*),  $\nu = 0.4$  - Fig. 5(*b*),  $\nu = 0.6$  - Fig. 5(*d*)] for which ADSs are not still available. As regards to the confidence

range  $\Delta_{ADS}$ , defined as

$$\Delta_{ADS} \triangleq \frac{PSL_{SUP}}{PSL_{INF}},\tag{16}$$

it slightly increases with PQ and shows a limited dependence on the aperture dimension (~ 4 dB in  $10^2 \le PQ \le 10^6$ ) (Fig. 6). Moreover,  $\Delta_{ADS}(\nu) = \Delta_{ADS}(1-\nu)$  and the minimum value of  $\Delta_{ADS}$  verifies for  $\nu = 0.5$  as it can be analytically derived.

Concerning available ADSs with  $\nu \neq 0.5$ , Figure 7 shows the behavior of the  $PSL_{opt}$  (and related bounds) of the array generated from the sequence  $\underline{\mathbf{D}}_3$  ( $\eta = 0.473$  and  $\nu = 0.485$ ) whose power pattern and elements arrangement are given in Fig. 7(*c*) and Fig. 7(*d*), respectively. Despite the small aperture  $(3\lambda \times 5\lambda)$ ,  $PSL_{opt}$  still lies in the range of values estimated by (13) [Figs. 7(*a*)-7(*b*)] and it appears to be significantly below the random estimates and comparable with the DS value at  $\eta = 1$ . It is also interesting to notice that the reference array derived from  $\underline{\mathbf{D}}_3$  allows one to determine several shifted array configurations with  $PSL\left\{\underline{\mathbf{D}}_3^{(\sigma_x,\sigma_y)}\right\} \leq PSL_{SUP}$  [Fig. 3(*b*)] as well as multiple arrays with  $PSL\left\{\underline{\mathbf{D}}_3^{(\sigma_x,\sigma_y)}\right\} \leq PSL_{max}$ .

Such a feature is not only limited to  $\underline{\mathbf{D}}_3$ , but it is a common property of ADS-based arrays as confirmed by the examples in Figs. 8-10 and concerned with larger apertures. Furthermore, it should be pointed out that more than one cyclic shift of the reference ADS sequence  $\underline{\mathbf{D}}_i$ (i = 4, ..., 7) gives an array pattern with  $PSL\left\{\underline{\mathbf{D}}_i^{(\sigma_x,\sigma_y)}\right\} = PSL_{opt}$  [Figs. 3(b), 8(a), 9(a), 10(a)]. Such considerations highlight that: (a) also through an exhaustive search, less than  $P \times Q$  evaluations are actually needed to identify the optimal ADS-based planar array; (b) a very limited number of evaluations is enough to synthesize an ADS array with a PSL value below that from random/random lattice distributions.

As far as the radiation patterns are concerned, Figures 7(c)-10(c) allow one to point out a further interesting property of ADS planar arrays. Unlike DSs, where  $|\mathbb{F}(k,l)|$  is a two-valued function [6], the unequal magnitudes of the samples of  $|\mathbb{F}(k,l)|$  (Fig. 2) lead to a non-uniform behavior of the array pattern outside the mainlobe region with some non-negligible variations of the sidelobes [see Figs. 7(c)-10(c)]. This can be exploited as an additional degree of freedom to be used in antenna synthesis. One efficient way to do that is to consider directive elements. As an example, let us consider the planar arrays synthesized from  $\underline{D}_4$  with isotropic or directive elements (e.g.,  $\frac{\lambda}{2}$  dipoles along the y axis). Figure 11(a) gives the PSL values for different shifts of the reference set. As it can be observed, the value of  $PSL_{opt}$  reduces ( $PSL_{opt}^{dir} = -23.66$  dB vs.  $PSL_{opt} = -21.79$  dB) thanks to the use of directive elements and, more interestingly, the optimal shift for the directive array is not equal to that with isotropic elements ( $\sigma_{x,opt}^{dir} = 5$ ,  $\sigma_{y,opt}^{dir} = 20$  vs.  $\sigma_{x,opt} = 7$ ,  $\sigma_{y,opt} = 5$ ). This is due to the following. One has to determine the shift generating the lowest lobes in the whole sidelobe region when dealing with the "isotropic" array [Fig. 11(*b*)]. Otherwise, the use of directive elements suggests to choose the  $\sigma_x$  and  $\sigma_y$  values with lowest sidelobes only near the mainlobe region [Fig. 11(*c*)] since the element factor "erases" the highest sidelobes far from the mainlobe region in the resulting antenna pattern [Fig. 11(*d*)].

The last section of the numerical validation is aimed at giving some indications on the performance of the ADS arrays versus those coming from state-of-the-art thinning techniques based on stochastic optimizers [7][23][5]. Since ADSs are not still available in correspondence with the thinning percentage of the test cases under analysis, the comparison cannot be considered fully fair, but it can be useful to suggest some guidelines for a fast and reliable choice of the most suitable synthesis procedure as well as on the achievable PSL results.

Figure 12 shows the PSL of the thinned arrays optimized with the methods in [5] [Fig. 12(*a*)], [7] [Fig. 12(*b*)], and [23] [Fig. 12(*c*)], respectively, along with the PSL bounds derived for the corresponding ADS-based arrays (i.e., only the values of  $PSL_{SUP}$  and  $PSL_{INF}$  since the ADS sequences, although theoretically existing, have not been yet determined). As it can be noticed, ADS-based arrays compare favourably in terms of PSL with global optimized designs since, even in the worst case (i.e.,  $\eta = 0.5$ ),  $PSL_{INF} < PSL^{glo} \le PSL_{SUP}$ .

### **5** Conclusions

In this paper, ADS have been considered for the design of thinned planar arrays. The research work is aimed at identifying the descriptive parameters of the ADS-based thinning technique as well as their effect on the array performances. Likewise the linear case [20], the objective of this study is to analytically define a "term of comparison" to help the array designer in identifying the synthesis approach allowing the optimal trade-off between computational resources and the achievable result in terms of PSL level. Towards this purpose, the performances of planar ADS-based arrays have been investigated and suitable bounds for the PSL have been determined thanks to a new formulation based on the properties of the two-dimensional DFT. Such an analysis has been validated by means of a large set of numerical experiments also aimed at comparing the predicted ADS performances with those of random distributions or stochastically optimized arrays. The obtained results have pointed out the following features of the ADS thinning technique:

- the *PSL* of the synthesized pattern is *a-priori* known when the *ADS* sequence is available in an explicit form, while suitable bounds are predictable, otherwise;
- because of the three-level autocorrelation function, *ADS* sequences guarantee additional degrees-of-freedom (compared to the *DS* case) to be profitably exploited (e.g., using directive elements) for fitting the design constraints;
- unlike iterative optimization or trial-and-test random synthesis techniques, the approach determines the array configuration just through simple shifts of a reference *ADS* sequence;
- thanks to the availability of rich repositories of *ADS*s also concerned with large *P* and *Q* indexes, wide apertures (impracticable for stochastic optimizers) can be dealt with;
- the use of *ADS* does not prevent their integration with optimization techniques, vice versa it could represent a way (to be explored in successive researches) to improve the convergence rate of iterative methods or for enabling stochastic searches in thinning large arrays by means of a suitable *ADS*-based initialization.

### Acknowledgements

A. Massa wishes to thank E. Vico for being and C. Pedrazzani for her continuous help and patience. Moreover, the authors are very grateful to M. Donelli for kindly providing some numerical results of computer simulations.

# Appendix

#### - Definition of the Mainlobe Region ${\it R}$

Starting from (12) as for planar DS arrays [6], it can be proved that the PSL of ADSbased arrays is close to the values of the samples of the array factor at  $u = u_{m+\frac{1}{2}} \triangleq \frac{m+1/2}{Px_0}$ ,  $v = v_{n+\frac{1}{2}} \triangleq \frac{n+1/2}{Qy_0}$ . By exploiting such an observation, it results that

$$PSL\left\{\underline{\mathbf{D}}^{(\sigma_{x},\sigma_{y})}\right\} \approx \frac{1}{K^{2}}max_{\left(u_{m+\frac{1}{2}},v_{n+\frac{1}{2}}\right)\notin R} \left|\frac{K(-1)^{m+n}}{PQ\sin\left[\frac{\pi}{P}\left(m+\frac{1}{2}\right)\right]\sin\left[\frac{\pi}{Q}\left(n+\frac{1}{2}\right)\right]} + \sum_{k=0,kl\neq 0}^{P-1}\sum_{l=0,kl\neq 0}^{Q-1} \frac{\mathbb{F}^{(\sigma_{x},\sigma_{y})}(-1)^{m+k+n+l}}{PQ\sin\left[\frac{\pi}{P}\left(m+k+\frac{1}{2}\right)\right]\sin\left[\frac{\pi}{Q}\left(n+l+\frac{1}{2}\right)\right]}\right|^{2}.$$
(17)

where the mainlobe region, R, is defined analogously to [6] as the visible region where the first term in (17) exceeds the magnitude of the second one. As regards to the first term, its magnitude is approximately equal to

$$\frac{K}{\pi^2 \left(m + \frac{1}{2}\right) \left(n + \frac{1}{2}\right)}$$

and for large values of P and Q. Moreover, the largest coefficients in the second term (i.e.,  $m + k + \frac{1}{2} = \pm 1$  and  $n + l + \frac{1}{2} = \pm 1$ ) of (17) are bounded by

$$\frac{4\max_{(k,l)\in\mathcal{H}_0}|\mathbb{F}^{(\sigma_x,\sigma_y)}|}{\pi^2}.$$

Thus, after simple manipulation, it is possible to show that R extends to the region limited by the following boundary inequality

$$u_{m+\frac{1}{2}}v_{n+\frac{1}{2}} \le \frac{K}{4PQx_{0}y_{0}\max_{(k,l)\in\mathcal{H}_{0}}|\mathbb{F}^{(\sigma_{x},\sigma_{y})}|}.$$
(18)

#### - Derivation of $PSL_{SUP}$ in (13)

With reference to discrete version of R,  $R_D$ ,

$$R_D \triangleq \left\{ (m,n) \in \mathbb{Z} \times \mathbb{Z} : \left( m + \frac{1}{2} \right) \left( n + \frac{1}{2} \right) \le \frac{K}{4 \max_{(k,l) \in \mathcal{H}_0} |W_{DFT}(k,l)|} \right\}$$
(19)

let us consider the following approximation of  $PSL^{opt} \{\underline{\mathbf{D}}\} = min_{(\sigma_x,\sigma_y)} \left[ PSL \left\{ \underline{\mathbf{D}}^{(\sigma_x,\sigma_y)} \right\} \right]$ 

$$PSL^{opt}\left\{\underline{\mathbf{D}}\right\} \lesssim min_{(\sigma_x,\sigma_y)} \left[\frac{max_{(k,l)\in\mathcal{H}_0}\left\{|\mathbb{F}^{(\sigma_x,\sigma_y)}\left(k,l\right)|^2\right\}}{K^2} \times \right]$$
(20)

$$\times \max_{(m,n)\notin R_D} \left| \sum_{k=0,kl\neq 0}^{P-1} \sum_{l=0,kl\neq 0}^{Q-1} \frac{e^{j\phi_{kl}^{(\sigma_x,\sigma_y)}}(-1)^{m+k+n+l}}{PQ\sin\left[\frac{\pi}{P}\left(m+k+\frac{1}{2}\right)\right]\sin\left[\frac{\pi}{Q}\left(n+l+\frac{1}{2}\right)\right]} \right|^2$$

where the complex coefficient  $\mathbb{F}^{(\sigma_x,\sigma_y)}$  has been expressed in terms of its amplitude,  $\sqrt{|\mathbb{F}^{(\sigma_x,\sigma_y)}(k,l)|^2}$ , and phase,  $\phi_{kl}^{(\sigma_x,\sigma_y)}$ .

It is worth pointing out that, likewise DSs,  $\phi_{kl}^{(\sigma_x,\sigma_y)}$  is not *a-priori* known as well as, unlike DSs, the term  $max_{(k,l)\in\mathcal{H}_0} \{|\mathbb{F}^{(\sigma_x,\sigma_y)}(k,l)|^2\}$  and they have to be estimated. Towards this end, by exploiting the circular correlation property of DFT [22], it is possible to state that

$$\left|\mathbb{F}^{(\sigma_x,\sigma_y)}(k,l)\right|^2 = DFT\left\{A_w^{ADS}(p,q)\right\} = K - \Lambda + PQ\Lambda\,\delta(k,l) + \Psi(k,l),\tag{21}$$

and to obtain the following relationship

$$max_{(k,l)\in\mathcal{H}_0}\left\{\left|\mathbb{F}^{(\sigma_x,\sigma_y)}\left(k,l\right)\right|^2\right\} = K - \Lambda + max_{(k,l)\in\mathcal{H}_0}\left\{\Psi(k,l)\right\}$$
(22)

where  $\Psi(k, l) \triangleq DFT \{ \psi(p, q) \}$  being  $\psi(p, q) \triangleq \sum_{r=1}^{PQ-1-t} \delta(p - l_{r,1}, q - l_{r,2}).$ Concerning the real-valued coefficients  $\Psi(k, l)$ , by applying the Parseval's theorem [22]

$$\frac{1}{PQ} \sum_{k=0}^{P-1} \sum_{l=0}^{Q-1} |\Psi(k,l)|^2 = \sum_{p=0}^{P-1} \sum_{q=0}^{Q-1} |\psi(p,q)|^2 = PQ - 1 - t,$$

and noticing that  $\sum_{p=0}^{P-1} \sum_{q=0}^{Q-1} |\psi(p,q)|^2 = PQ - 1 - t$  and  $\Psi(0,0) = PQ - 1 - t$ , the following holds true

$$\sum_{k=0,kl\neq 0}^{P-1} \sum_{l=0,kl\neq 0}^{Q-1} |\Psi(k,l)|^2 = (t+1)(PQ-1-t).$$
(23)

Therefore, since

$$\max_{(k,l)\in\mathcal{H}_0} \left\{ \Psi(k,l) \right\} \le \max_{(k,l)\in\mathcal{H}_0} \left\{ |\Psi(k,l)| \right\} \le \sqrt{(t+1)(PQ-1-t)}$$
(24)

and substituting (22) in (20), we obtain

$$SL^{opt} \{\underline{\mathbf{D}}\} \lesssim \left(K - \Lambda + \sqrt{(t+1)(PQ - 1 - t)}\right) \times$$

$$\times min_{(\sigma_x, \sigma_y)} \left[\frac{max_{(m,n)\notin R_D} \left|\sum_{k=0,kl\neq 0}^{P-1} \sum_{l=0,kl\neq 0}^{Q-1} \frac{e^{j\phi_{kl}^{(\sigma_x,\sigma_y)}}(-1)^{m+k+n+l}}{PQ\sin[\frac{\pi}{P}(m+k+\frac{1}{2})]\sin[\frac{\pi}{Q}(n+l+\frac{1}{2})]}\right|^2}{K^2}\right].$$
(25)

As regards the phase terms  $\phi_{kl}^{(\sigma_x,\sigma_y)}$ , the analysis is carried out as in [6] in order to give an estimate of the *PSL*. More specifically, although the phase terms  $\phi_{kl}^{(\sigma_x,\sigma_y)}$  are deterministic quantities, they are dealt with as independent identically distributed (i.i.d.) uniform random variables. Under this assumption, (25) can be expressed as

$$PSL^{opt}\left\{\underline{\mathbf{D}}\right\} \lesssim \frac{\left(K - \Lambda + \sqrt{(t+1)(PQ - 1 - t)}\right)\Gamma^{opt}}{K^2}$$
(26)

where for large P and Q

$$\Gamma^{opt} \approx \min_{(\sigma_x, \sigma_y)} \left\{ \max\left[U_i; i = 1, ..., \Pi\right] \right\}$$
(27)

 $U_{i} \triangleq \left| \sum_{k=-\infty}^{\infty} \sum_{l=-\infty}^{\infty} \frac{e^{j\phi_{kl}^{(\sigma_{x},\sigma_{y})}}}{\pi^{2}(k+\frac{1}{2})(l+\frac{1}{2})} \right|^{2}, i = 1, ..., \Pi, \text{ being i.i.d. random variables and } \Pi \text{ is the cardinality of } R_{D} (\approx PQ).$  Since the statistics of  $\Gamma^{opt}$  are not available in closed form, Monte Carlo simulations have been performed to provide an approximation of its mean value  $E \{\Gamma^{opt}\}$ 

$$E\left\{\Gamma^{opt}\right\} \approx -0.1 + 1.5 \log_{10}(PQ).$$
 (28)

By substituting (28) in (26), the upper bound  $PSL_{SUP}$  is finally obtained.

#### - Derivation of $PSL_{INF}$ in (13)

By sampling (12) at  $(u = \frac{s}{Pd_x}, v = \frac{t}{Qd_y})$ , s = 0, ..., P - 1, t = 0, ..., Q - 1, it can be easily shown that

$$PSL^{opt}\left\{\underline{\mathbf{D}}\right\} \ge PSL\left\{\underline{\mathbf{D}}^{(\sigma_x,\sigma_y)}\right\}\Big|_{u=\frac{s}{Pd_x}, v=\frac{t}{Qd_y}} =$$
(29)

$$=\frac{\max_{(s,t)\in\mathcal{H}_0}\left|\sum_{k=0}^{P-1}\sum_{l=0}^{Q-1}\mathbb{F}^{(\sigma_x,\sigma_y)}(k,l)\frac{\sin[\pi(s+k)]}{P\sin\left[\frac{\pi(s+k)}{P}\right]}\frac{\sin[\pi(t+l)]}{Q\sin\left[\frac{\pi(t+l)}{Q}\right]}\right|^2}{K^2}}{K^2}=\frac{\max_{(s,t)\in\mathcal{H}_0}\left|\mathbb{F}^{(\sigma_x,\sigma_y)}(s,t)\right|^2}{K^2}$$

By substituting (22) in (29), and observing that

$$max_{(k,l)\in\mathcal{H}_0} \{\Psi(k,l)\} \ge -\sqrt{\frac{(t+1)(PQ-1-t)}{PQ-1}}$$
 (30)

it turns out

$$PSL^{opt}\left\{\underline{\mathbf{D}}\right\} \ge \frac{K - \Lambda - \sqrt{\frac{(t+1)(PQ-1-t)}{PQ-1}}}{K^2} \triangleq PSL_{INF}.$$
(31)

#### - Derivation of $PSL_{max}$ in (13)

With reference to (20), let us assume that the  $ADS \underline{D}$  at hand is known. Thus,

$$\Omega\left\{\underline{\mathbf{D}}\right\} \triangleq \max_{(k,l)\in\mathcal{H}_{0}}\left|\mathbb{F}^{(\sigma_{x},\sigma_{y})}\left(k,l\right)\right|^{2}$$
(32)

is now a known quantity. By substituting (32) in (20), we obtain

$$PSL^{opt}\left\{\underline{\mathbf{D}}\right\} \lesssim \Omega\left\{\underline{\mathbf{D}}\right\} \min_{(\sigma_x, \sigma_y)} \left[ max_{(m,n)\notin R_D} \left| \frac{\sum_{k=0,kl\neq 0}^{P-1} \sum_{l=0,kl\neq 0}^{Q-1} \frac{e^{j\phi_{kl}^{(\sigma_x, \sigma_y)}}(-1)^{m+k+n+l}}{PQ\sin\left[\frac{\pi}{P}\left(m+k+\frac{1}{2}\right)\right]\sin\left[\frac{\pi}{Q}\left(n+l+\frac{1}{2}\right)\right]}{K^2} \right|^2 \right]$$

$$(33)$$

As regards to the phase terms  $\phi_{kl}^{(\sigma_x,\sigma_y)}$ , let us consider the same procedure used for deriving  $PSL_{SUP}$  and let us rewrite (33) as follows

$$PSL^{opt}\left\{\underline{\mathbf{D}}\right\} \lesssim \frac{\Omega\left\{\underline{\mathbf{D}}\right\}\Gamma^{opt}}{K^2}$$
(34)

where  $\Gamma^{opt}$  is successively approximated with its mean value (28) to obtain  $PSL_{max}$ .

- Derivation of  $PSL_{min}$  in (13)

Analogously to the derivation of  $PSL_{max}$ , a lower bound for  $PSL_{opt}$  when  $\underline{D}$  is known can be obtained starting from (29) and employing (32):

$$PSL^{opt}\left\{\underline{\mathbf{D}}\right\} \ge \frac{max_{(s,t)\in\mathcal{H}_0} \left|\mathbb{F}^{(\sigma_x,\sigma_y)}\left(s,t\right)\right|^2}{K^2} = \frac{\Omega\left\{\underline{\mathbf{D}}\right\}}{K^2} \triangleq PSL_{min}.$$
(35)

### References

- [1] C. A. Balanis, Antenna Theory: Analysis and Design, 2nd ed. New York: Wiley, 1997.
- Y. T. Lo, "A mathematical theory of antenna arrays with randomly spaced elements," *IEEE Trans. Antennas Propag.*, vol. 12, no. 3, pp. 257-268, May 1964.
- [3] B. Steinberg, "The peak sidelobe of the phased array having randomly located elements," *IEEE Trans. Antennas Propag.*, vol. 20, no. 2, pp. 129-136, Mar. 1972.
- [4] D. G. Leeper, "Thinned periodic antenna arrays with improved peak sidelobe level control," U.S. Patent 4071848, Jan. 31, 1978.
- [5] R. L. Haupt, "Thinned arrays using genetic algorithms," *IEEE Trans. Antennas Propag.*, vol. 42, no. 7, pp. 993-999, Jul. 1994.
- [6] D. G. Leeper, "Isophoric arrays massively thinned phased arrays with well-controlled sidelobes," *IEEE Trans. Antennas Propag.*, vol. 47, no. 12, pp. 1825-1835, Dec 1999.
- [7] S. Caorsi, A. Lommi, A. Massa, and M. Pastorino, "Peak sidelobe reduction with a hybrid approach based on GAs and difference sets," *IEEE Trans. Antennas Propag.*, vol. 52, no. 4, pp. 1116-1121, Apr. 2004.
- [8] R. L. Haupt and D. H. Werner, *Genetic algorithms in electromagnetics*. Hoboken, NJ: Wiley, 2007.

- [9] B. Steinberg, "Comparison between the peak sidelobe of the random array and algorithmically designed aperiodic arrays," *IEEE Trans. Antennas Propag.*, vol. 21, no. 3, pp. 366-370, May 1973.
- [10] S. Holm, B. Elgetun, and G. Dahl, "Properties of the beampattern of weight- and layout-optimized sparse arrays," *IEEE Trans. Ultrason., Ferroelectr., Freq. Control*, vol. 44, no. 5, pp. 983-991, Sep. 1997.
- [11] T. G. Spence, D. H. Werner, "Thinning of aperiodic antenna arrays for low side-lobe levels and broadband operation using genetic algorithms," IEEE Antennas and Propagation Society International Symposium 2006, pp. 2059-2062, 9-14 July 2006.
- [12] A. Trucco and V. Murino, "Stochastic optimization of linear sparse arrays," *IEEE J. Ocean Eng.*, vol. 24, no. 3, pp. 291-299, Jul. 1999.
- [13] A. Trucco, "Thinning and weighting of large planar arrays by simulated annealing," *IEEE Trans. Ultrason., Ferroelectr., Freq. Control*, vol. 46, no. 2, pp. 347-355, Mar. 1999.
- [14] L. E. Kopilovich, "Square array antennas based on hadamard difference sets," *IEEE Trans. Antennas Propag.*, vol. 56, no. 1, pp. 263-266, Jan. 2008.
- [15] La Jolla Cyclic Difference Set Repository (http://www.ccrwest.org/diffsets.html).
- [16] C. Ding, T. Helleseth, and K. Y. Lam, "Several classes of binary sequences with three-level autocorrelation," *IEEE Trans. Inf. Theory*, vol. 45, no. 7, pp. 2606-2612, Nov. 1999.
- [17] K. T. Arasu, C. Ding, T. Helleseth, P. V. Kumar, and H. M. Martinsen, "Almost difference sets and their sequences with optimal autocorrelation," *IEEE Trans. Inf. Theory*, vol. 47, no. 7, pp. 2934-2943, Nov 2001.
- [18] Y. Zhang, J. G. Lei, and S. P. Zhang, "A new family of almost difference sets and some necessary conditions," *IEEE Trans. Inf. Theory*, vol. 52, no. 5, pp. 2052-2061, May 2006.
- [19] ELEDIA Almost Difference Set Repository (http://www.ing.unitn.it/~eledia/html/).

- [20] G. Oliveri, M. Donelli, and A. Massa, "Linear array thinning exploiting almost difference sets," *IEEE Trans. Antennas Propag.*, revised.
- [21] M. I. Kargapolov and J. I. Merzljako, *Fundamentals of the Theory of Groups*. New York: Springer-Verlag, 1979.
- [22] J. G. Proakis and D. G. Manolakis, *Digital Signal Processing: Principles, Algorithms, and Applications*, 3nd ed. London: Prentice Hall, 1996.
- [23] M. Donelli, A. Martini, and A. Massa, "A hybrid approach based on PSO and Hadamard difference sets for the synthesis of square thinned arrays," *IEEE Trans. Antennas Propag.* (*in press*).

#### **FIGURE CAPTIONS**

- Figure 1. Autocorrelation functions and associated binary sequences of the ADSs in Tab.
  I: (a)(d) <u>D</u><sub>1</sub>, (b)(e) <u>D</u><sub>2</sub>, and (c)(f) <u>D</u><sub>3</sub>.
- Figure 2. Plot of the normalized array factor derived from <u>D</u><sub>i</sub> = <u>D</u><sub>i</sub><sup>(σ<sub>x</sub>=0, σ<sub>y</sub>=0)</sup> and associated |F(k, l)| values: (a) i = 1 and (b) i = 2.
- Figure 3. PSL values of the ADS-based planar arrays derived from the sequences <u>D</u><sub>3</sub><sup>(σ<sub>x</sub>, σ<sub>y</sub>)</sup>, σ<sub>x</sub> = 0, ..., P − 1, σ<sub>y</sub> = 0, ..., Q − 1 (a) and PSL bounds (b). Number of elements: P × Q = 7 × 11 - Aperture size: 3λ × 5λ.
- Figure 4. Numerical Validation. Plots of the PSL bounds of ADS-based planar arrays, the estimator of the PSL of random (RND) and random lattice (RNL) arrays, and values of the PSL of DS-based finite arrays versus η when ν = 0.5 and (a) PQ = 10<sup>2</sup>, (b) PQ = 10<sup>4</sup>, (c) PQ = 10<sup>6</sup>.
- Figure 5. Numerical Validation. Plots of the PSL bounds of ADS-based planar arrays and the estimators of the PSL of random (RND) and random lattice (RNL) arrays versus the array aperture, PQ, when η = 0.5 and (a) ν = 0.3, (b) ν = 0.4, (c) ν = 0.5, (d) ν = 0.6.
- Figure 6. Numerical Validation. Plots of  $\Delta_{ADS}$  versus the array aperture, PQ, when  $\eta = 0.5$  and in correspondence with different thinning values [ $\nu \in [0, 1]$ ].
- Figure 7. Numerical Validation Planar Array <u>D</u><sub>3</sub><sup>opt</sup> [Number of elements: P × Q = 7 × 11 Aperture size: 3λ × 5λ]. Plots of the PSL bounds versus (a) the array aperture PQ [ν = 0.4805, η = 0.4736] and (b) η [PQ = 77, ν = 0.4805]. Plot of the normalized array factor (c) generated from the <u>D</u><sub>3</sub><sup>opt</sup>-based array arrangement (d).
- Figure 8. Numerical Validation Planar Array <u>D</u><sup>opt</sup><sub>4</sub> [Number of elements: P × Q = 23 × 23 Aperture size: 11λ × 11λ]. PSL values of the ADS-based arrays derived from the sequences <u>D</u><sup>(σx, σy)</sup><sub>4</sub>, σ<sub>x</sub> = 0, ..., P − 1, σ<sub>y</sub> = 0, ..., Q − 1 (a). Plots of the PSL bounds

versus (b)  $\eta$  [PQ = 529,  $\nu = 0.5$ ]. Plot of the normalized array factor (c) generated from the  $\underline{\mathbf{D}}_{4}^{opt}$ -based array arrangement (d).

- Figure 9. Numerical Validation Planar Array D<sub>5</sub><sup>opt</sup> [Number of elements: P × Q = 73 × 23 Aperture size: 36λ × 36λ]. PSL values of the ADS-based arrays derived from the sequences D<sub>5</sub><sup>(σ<sub>x</sub>, σ<sub>y</sub>)</sup>, σ<sub>x</sub> = 0, ..., P − 1, σ<sub>y</sub> = 0, ..., Q − 1 (a). Plots of the PSL bounds versus (b) η [PQ = 5329, ν = 0.5]. Plot of the normalized array factor (c) generated from the D<sub>5</sub><sup>opt</sup>-based array arrangement (d).
- Figure 10. Numerical Validation Planar Array <u>D</u><sub>6</sub><sup>opt</sup> [Number of elements: P × Q = 199 × 199 Aperture size: 99λ × 99λ]. PSL values of the ADS-based arrays derived from the sequences <u>D</u><sub>6</sub><sup>(σ<sub>x</sub>, σ<sub>y</sub>)</sup>, σ<sub>x</sub> = 0, ..., P − 1, σ<sub>y</sub> = 0, ..., Q − 1 (a). Plots of the PSL bounds versus (b) η [PQ = 39601, ν = 0.5]. Plot of the normalized array factor (c) generated from the <u>D</u><sub>6</sub><sup>opt</sup>-based array arrangement (d).
- Figure 11. Numerical Validation Non-Isotropic elements. PSL values of the ADS-based arrays generated from the sequences <u>D</u><sup>(σ<sub>x</sub>, σ<sub>y</sub>)</sup>, σ<sub>x</sub> = 0, ..., P 1, σ<sub>y</sub> = 0, ..., Q 1
  (a). Normalized array patterns of the arrays generated from the sequences (b) <u>D</u><sup>opt</sup><sub>4</sub> = <u>D</u><sup>(7,5)</sup><sub>4</sub> with isotropic elements, <u>D</u><sup>(5,20)</sup><sub>4</sub> with isotropic (c) and directive elements (d).
- Figure 12. Comparative Assessment. Plots of the PSL bounds versus the array aperture, PQ, when η = 0.5 and for (a) ν = 0.54 [5], (b) ν = 0.507 [7], (c) ν = 0.48 [23], (d) ν = 0.44 [23].

## **TABLE CAPTIONS**

• Table I. Examples of *ADS*s and descriptive parameters.













(*e*)





(f)

Figure 1 - G. Oliveri et al., "ADS-Based Guidelines for ..."



(*a*)



(*b*)

Figure 2 - G. Oliveri et al., "ADS-Based Guidelines for ..."





(*b*)

Figure 3 - G. Oliveri et al., "ADS-Based Guidelines for ..."











Figure 4 - G. Oliveri et al., "ADS-Based Guidelines for ..."







Figure 6 - G. Oliveri et al., "ADS-Based Guidelines for ..."



value [dB]



(*c*)



Figure 7 - G. Oliveri et al., "ADS-Based Guidelines for ..."

(d)



value [dB]

Datie



(*c*)



Figure 8 - G. Oliveri et al., "ADS-Based Guidelines for ..."



(*c*)

Figure 9 - G. Oliveri et al., "ADS-Based Guidelines for ..."

30

*(d)* 





(d)





-15

-25

-0.5

0 u 0.5

-1



*(b)* 

![](_page_33_Figure_4.jpeg)

(*c*)

0.5 0 v

-1

(d)

![](_page_34_Figure_0.jpeg)

![](_page_34_Figure_1.jpeg)

ADS	Р	Q	K	Λ	t	$\nu \triangleq \frac{K}{PQ}$	$\eta \triangleq \frac{t}{PQ-1}$	Reference
$\underline{\mathbf{D}}_1$	4	7	15	7	6	$\approx 0.5357$	$\approx 0.22$	[18]
$\underline{\mathbf{D}}_2$	7	7	25	12	24	$\approx 0.5102$	0.5	[18]
$\underline{\mathbf{D}}_3$	7	11	37	17	36	$\approx 0.4805$	$\approx 0.4736$	[18]
$\underline{\mathbf{D}}_4$	23	23	265	132	264	$\approx 0.5$	0.5	[19]
$\underline{\mathbf{D}}_5$	73	73	2665	1332	2664	$\approx 0.5$	0.5	[19]
$\underline{\mathbf{D}}_{6}$	199	199	19801	9900	19800	$\approx 0.5$	0.5	[19]

Table I - G. Oliveri et al., "ADS-Based Guidelines for ..."