

On the Design of Modular Reflecting *EM* Skins for Enhanced Urban Wireless Coverage

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Abstract—The design of modular, passive, and static artificial metasurfaces to be used as electromagnetic skins (*EMS*s) of buildings for improving the coverage in urban millimeter-wave communication scenarios is addressed. Towards this end, an *ad-hoc* design strategy is presented to determine optimal trade-off implementative solutions that assure a suitable coverage of the areas of interest, where the signal from the base station is too weak, with the minimum complexity. More specifically, the admissible surface in the building facade is first partitioned into tiles, which are the minimum-size elements of the artificial coating (i.e., the building block of an *EMS*). Then, the search for the optimal *EMS* layout (i.e., the minimum number and the positions of the tiles to be installed) is carried out with a binary multi-objective optimization method. Representative numerical results are reported and discussed to point out the features and the potentialities of the *EMS* solution in the smart electromagnetic environment (*SEME*) as well as the effectiveness of the proposed design method.

Index Terms—Smart EM Environment, Artificial Materials, Mobile Communications, Millimeter-wave, Multi-Objective Optimization.

I. INTRODUCTION

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STARTING from the first generation of cellular networks, back when the communications were analog and the portable devices heavy and cumbersome, there has always been a continuous technological push towards higher data rates, which are a mandatory requirement with the introduction of smart-phones and the sharing of multimedia contents now enabled by the fast and reliable data streams of 4G and 5G communication networks [1]-[3]. Regardless of the throughput of modern communication networks, the data traffic is expected to further increase in the next years because of the massive proliferation of wireless devices and systems (e.g., smart-phones, tablets, internet-of-things (*IoT*) sensors, and robots) as well as the introduction of novel applications including, for instance, the autonomous driving [4], the tactile internet [5], and the remote control of robots [6]. All these applications will unavoidably require improved coverage (i.e., a higher level of the *EM* signal in the coverage area) and better quality-of-service (*QoS*) for mobile users/devices as well as wireless links characterized by lower latency and higher throughput/resiliency [7][8]. Towards this end, future mobile communication networks will have to assure more and more reliable and ubiquitous connections, everywhere and anytime, as never seen before. However, the standard solutions chosen by the operators (i.e., installing more base stations (*BS*s), transmitting more power, or using new frequency bands) are no longer applicable because of the too high power consumption due to the foreseen explosion of the traffic needs as well as the spectrum congestion [9]. Moreover, the obstacles/scatterers in the environment cannot be neglected due to the increase in the operation frequencies (e.g., millimeter-waves in 5G [10]) so that the no line-of-sight (*NLOS*) condition has to be taken into account since the design of the wireless network architecture. A possible countermeasure to these issues and challenges is the “implementation” of the so-called smart electromagnetic environment (*SEME*) [11], where the objects and the scatterers within the environment are considered, unlike the past, as enablers [12] of the electromagnetic (*EM*) propagation and not impairments. The environment is thus exploited as an additional degree-of-freedom (*DoF*) for tailoring the propagation of the *EM* waves and to enhance the signal strength in the “blind spots”, namely the zones where the signal from the base station is too weak to support a desired throughput for users’ applications [12]. By using artificial materials (e.g., engineered materials or metamaterials [13]) on the building facades or integrated within panels along the streets, the propagation of the *EM* waves in complex urban scenarios is controlled to fit *QoS* requirements and coverage targets. Passive (i.e., without

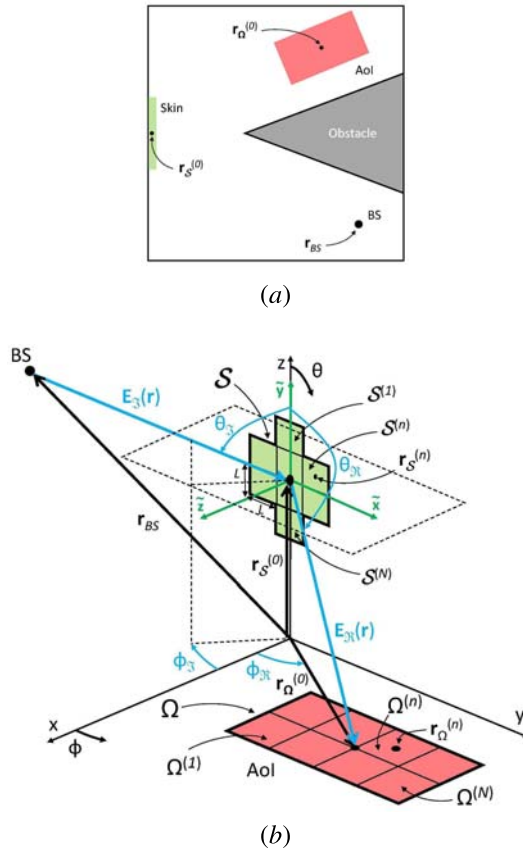


Figure 1. Problem geometry. Illustrative sketches of the wireless communication scenario: (a) top view and (b) detailed zoom.

active amplifiers) reconfigurable metasurfaces, which behave like intelligent reflecting surfaces (IRSs) [14]–[20] thanks to simple electronic devices such as radio-frequency switches [21], have been successfully installed, but they will not be massively deployed until their cost is significantly reduced.

An opposite strategy still profitably exploits the objects in the environment, without introducing additional materials, but optimizing the excitations of the array elements of the BS to generate a desired EM field distribution in the desired spots [22]. Although there are no additional costs and an installation of new hardware is not required, the opportunistic use of the BS array offers to the designer a limited number of DoFs and an accurate knowledge of the surrounding scenario is also needed to assure a suitable/stable QoS.

Differently, this paper is concerned with the instance of the SEME vision aimed at providing a doable large-scale solution suitable for the mass production. More specifically, it proposes a method for the design of modular, passive, and static metasurfaces to build effective and low-cost EMSs. In telecommunication engineering, the term EMS refers to a device conformal to the external surface of the object, where it is installed, that offers a set of functionalities related to the sensing and the manipulation of the EM waves [23]. A typical deployment for an EMS is over the facades of a building, which are strategic assets in a urban scenario for redirecting the impinging EM field towards areas where the signal would otherwise be weak. Starting from the selection/definition of the

maximal area available on a building facade, the support of the EMS is discretized into tiles, which are the elementary building blocks of an EMS. The arrangement of the tiles on the admissible surface of the facade is then optimized with a multi-objective global optimization strategy based on a binary implementation of the Non-Dominated Sorting Genetic Algorithm II (NSGA-II) [24] to yield the optimal trade-off solutions between the best coverage of the zones of interest and the minimum number of tiles.

To the best of the authors' knowledge, the main innovative contributions of this research work include (i) the description, the statement, and the mathematical formalization of a novel design problem within the SEME framework, (ii) the development of a customized design strategy for selecting the minimum number of tiles of the EMS that assures the coverage of the regions of interest, and (iii) the synthesis of innovative tiled EMSs to be embedded in the facade of buildings for improving the EM coverage within urban millimeter-wave communication scenarios.

The rest of the paper is organized as follows. The synthesis problem of the modular, passive, and static EMS is mathematically described and formulated in Sect. 2, while the optimization-based design method is presented in Sect. 3. Section 4 deals with the validation and the numerical assessment of the proposed concepts and synthesis method by considering realistic urban scenarios served by a millimeter-wave band of 5G systems. Eventually, some conclusions and final remarks are drawn (Sect. 5).

II. MATHEMATICAL FORMULATION

Let us consider an urban scenario [Fig. 1(a)] where a BS serves the terminals and the devices in the surrounding environment to implement the communication network by providing wireless services to the users with a suitable QoS. Because of the presence and the configuration of the buildings, the signal from the BS is absent or too weak to guarantee a connection to the network or a sufficient throughput in some areas, $\{\Omega_b; b = 1, \dots, B\}$, of the scenario at hand. In order to increase the signal strength in these “blind spots”, the installation of artificial EMSs is considered.

For the sake of formulation simplicity, the case of a single “blind spot” (i.e., $B = 1, \Omega \leftarrow \Omega_b$), where there is neither direct nor reflected signal from the BS is taken into account. Accordingly, only the effect of the EM field reflected from the EMS without considering the presence of the walls and the design of a single EMS on a facade of a building are considered in this work. However, it is worthwhile pointing out that both the theoretical description and the proposed design method can be straightforwardly extended to the case with multiple skins over multiple buildings to cover multiple blind spots. With reference to such a benchmark, an area S on the external wall of a selected building is assumed to be located on the $y-z$ plane, subdivided into N square tiles, $\{S^{(n)}; n = 1, \dots, N\}$ ($\bigcup_{n=1}^N S^{(n)} = S$), of equal size ΔS ($\Delta S = L \times L$, L being the side length of each tile), and centered at the positions $\mathbf{r}_S^{(n)} = y_S^{(n)}\hat{\mathbf{y}} + z_S^{(n)}\hat{\mathbf{z}}$ ($n = 1, \dots, N$) [Fig. 1(b)], while the BS is located at $\mathbf{r}_{BS} = x_{BS}\hat{\mathbf{x}} + y_{BS}\hat{\mathbf{y}} + z_{BS}\hat{\mathbf{z}}$ in the far-field of S .

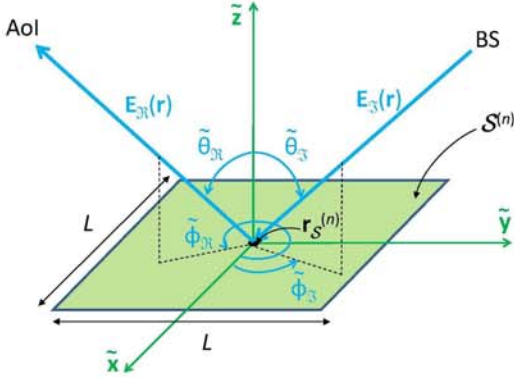


Figure 2. Problem geometry. Graphical representation of the EMS local coordinate system.

Without loss of generality, the EM wave generated from the BS and impinging on the EMS is modeled as a monochromatic plane wave at the working frequency f with electric field [25][26]

$$\mathbf{E}_3(\mathbf{r}) \triangleq E_3 \hat{\mathbf{e}}_3 e^{-j\mathbf{k}_3 \cdot (\mathbf{r} - \mathbf{r}_S^{(0)})} \quad (1)$$

where E_3 is the complex-valued wave amplitude, $\hat{\mathbf{e}}_3$ is the complex polarization vector, and \mathbf{k}_3 ($\mathbf{k}_3 \triangleq -k[\sin\theta_3 \cos\phi_3 \hat{\mathbf{x}} + \sin\theta_3 \sin\phi_3 \hat{\mathbf{y}} + \cos\theta_3 \hat{\mathbf{z}}]$) is the incident wave vector, k being the free-space wavenumber ($k \triangleq \frac{2\pi}{\lambda}$, λ being the wavelength at f), while $(\theta_3^{(0)}, \phi_3^{(0)})$ is the direction of arrival of the incident wave from the BS to the center of \mathcal{S} , $\mathbf{r}_S^{(0)}$ ($\mathbf{r}_S^{(0)} = y_S^{(0)} \hat{\mathbf{y}} + z_S^{(0)} \hat{\mathbf{z}}$) [Fig. 1(b)], being $\theta_3^{(n)} \Big|_{n=0} = \arccos\left(\frac{z_{BS} - z_S^{(n)}}{|\mathbf{r}_{BS} - \mathbf{r}_S^{(n)}|}\right)$ and $\phi_3^{(n)} \Big|_{n=0} = \arctan\left(\frac{y_{BS} - y_S^{(n)}}{x_{BS} - x_S^{(n)}}\right) \Big|_{n=0} = \arctan\left(\frac{y_{BS} - y_S^{(n)}}{x_{BS}}\right) \Big|_{n=0}$ since $x_S^{(0)} = 0$.

Each tile $\mathcal{S}^{(n)}$ ($n = 1, \dots, N$) of the EMS is illuminated from the direction $(\theta_3^{(n)}, \phi_3^{(n)})$ and it reflects the impinging wave towards a different direction $(\theta_R^{(n)}, \phi_R^{(n)})$ [$\theta_R^{(n)} = \arccos\left(\frac{z_\Omega^{(n)} - z_S^{(n)}}{|\mathbf{r}_\Omega^{(n)} - \mathbf{r}_S^{(n)}|}\right)$, $\phi_R^{(n)} = \arctan\left(\frac{y_\Omega^{(n)} - y_S^{(n)}}{x_\Omega^{(n)} - x_S^{(n)}}\right)$] ($n = 1, \dots, N$) by focusing the reflected beam in the point $\mathbf{r}_\Omega^{(n)}$ ($\mathbf{r}_\Omega^{(n)} = x_\Omega^{(n)} \hat{\mathbf{x}} + y_\Omega^{(n)} \hat{\mathbf{y}} + z_\Omega^{(n)} \hat{\mathbf{z}}$, $n = 1, \dots, N$) of the coverage region Ω , which is also called area of interest (AoI). More in detail, the AoI is assumed parallel to the $x - y$ plane (i.e., $z_\Omega^{(n)} = z_\Omega$, $\forall n$), centered at $\mathbf{r}_\Omega^{(0)}$, and discretized into N partitions, $\{\Omega^{(n)}; n = 1, \dots, N\}$ (i.e., $\bigcup_{n=1}^N \Omega^{(n)} = \Omega$), with equal dimensions $\Delta\Omega^{(n)} = \Delta\Omega$, having barycenters at $\mathbf{r}_\Omega^{(n)}$ ($n = 1, \dots, N$) along the directions of the reradiation angles [Fig. 1(b)].

By neglecting the polarization and the reflection losses, the far-field expression of the electric field reflected from the n -th ($n = 1, \dots, N$) tile $\mathcal{S}^{(n)}$ in a generic point $\tilde{\mathbf{r}}$ of the local coordinate system¹ (Fig. 2), $\mathbf{E}_R^{(n)}(\tilde{\mathbf{r}}) = E_R^{(n)}(\tilde{\mathbf{r}}) \hat{\mathbf{e}}_R$, is given

by the following closed-form expression [29]

$$E_R^{(n)}(\tilde{\mathbf{r}}) \simeq -jk\eta \frac{e^{-jk(d_3^{(n)} + d_R^{(n)})}}{4\pi d_3^{(n)} d_R^{(n)}} L^2 \times (\cos\tilde{\theta}_3^{(n)} + \cos\tilde{\theta}_R^{(n)}) \text{sinc}(kL\mathcal{D}_{\tilde{x}}) \text{sinc}(kL\mathcal{D}_{\tilde{y}}) \times e^{-j(\varphi_3 + \varphi_R)} \quad (2)$$

where η is the free-space impedance, while $d_3^{(n)}$ and $d_R^{(n)}$ are the distances travelled by the incident wave from the BS to the barycenter of the n -th ($n = 1, \dots, N$) tile ($d_3^{(n)} = |\mathbf{r}_{BS} - \mathbf{r}_S^{(n)}|$) and by the reflected wave from the barycenter of the n -th ($n = 1, \dots, N$) tile to the point $\mathbf{r}_\Omega^{(n)}$ where the peak of the reflected beam, generated from the same n -th tile, is directed ($d_R^{(n)} = |\mathbf{r}_S^{(n)} - \mathbf{r}_\Omega^{(n)}|$), respectively. Moreover, φ_3 is the phase associated to the modulation of the signal generated by the BS, φ_R is the phase term that can be engineered with the design of the surface, while $\mathcal{D}_{\tilde{x}} = \sin\tilde{\theta} \cos\tilde{\phi} - \sin\tilde{\theta}_R \cos\tilde{\phi}_R$ and $\mathcal{D}_{\tilde{y}} = \sin\tilde{\theta} \sin\tilde{\phi} - \sin\tilde{\theta}_R \sin\tilde{\phi}_R$. To improve the coverage within the AoI by guaranteeing a suitable intensity of the signal from the BS by means of the reflection from the EMS, the following synthesis problem is formulated:

Modular Reflecting EM Skin (MREMS) Design

Problem - Given an admissible skin surface \mathcal{S} discretized into N square tiles, $\{\mathcal{S}^{(n)}; n = 1, \dots, N\}$, and an AoI Ω , select the minimum number of tiles M (i.e., $M \leq N$), which reflect the EM wave from the BS towards the corresponding focusing points, $\{\mathbf{r}_\Omega^{(n)}; n = 1, \dots, M\}$, within the AoI ($\mathbf{r}_\Omega^{(n)} \in \Omega$), so that the power collected by a receiver at the position \mathbf{r}_u of the AoI ($\mathbf{r}_u \in \Omega$), $\mathcal{P}_R(\mathbf{r}_u)$ ($\mathcal{P}_R(\mathbf{r}) \triangleq \sum_{m=1}^M |E_R^{(m)}(\mathbf{r})|^2$), fulfils the condition

$$\mathcal{P}_R(\mathbf{r}_u) \geq \mathcal{P}_{th} \quad (3)$$

where \mathcal{P}_{th} is a user-defined coverage threshold ($\mathcal{P}_{th} \geq \mathcal{P}_{bts}$, \mathcal{P}_{bts} being the minimum level for a wireless connection).

III. MREMS SYNTHESIS METHOD

The MREMS Design Problem is addressed through a multi-objective optimization strategy based on a binary implementation of the NSGA-II [24]. Towards this end, the presence/absence of a tile on the final EMS layout \mathcal{S}_{opt} is mathematically modeled by means of a binary variable, t_n ($t_n \in \{0, 1\}$) ($n = 1, \dots, N$). If $t_n = 1$, then the n -th ($n = 1, \dots, N$) tile $\mathcal{S}^{(n)}$ is present on the facade of the building and it contributes to the reflection of the EM wave towards the AoI. Otherwise (i.e., $t_n = 0$), the n -th ($n = 1, \dots, N$) tile $\mathcal{S}^{(n)}$ is not installed on the external wall of the building, which maintains its original scattering properties without contributing to the enhancement of the signal in Ω . Accordingly, an admissible layout of the EMS (i.e., an arrangement of tiles over the available surface \mathcal{S}) is univocally described by the binary vector \mathbf{T} ($\mathbf{T} \triangleq \{t_n; n = 1, \dots, N\}$).

In order to determine the final structure of the EMS, \mathcal{S}_{opt} , namely the best subset of M tiles among the N admissible

¹The following relationships between the reference and the local coordinate systems hold true: $\tilde{x} = y$, $\tilde{y} = z - z_S^{(n)}$, and $\tilde{z} = x$.

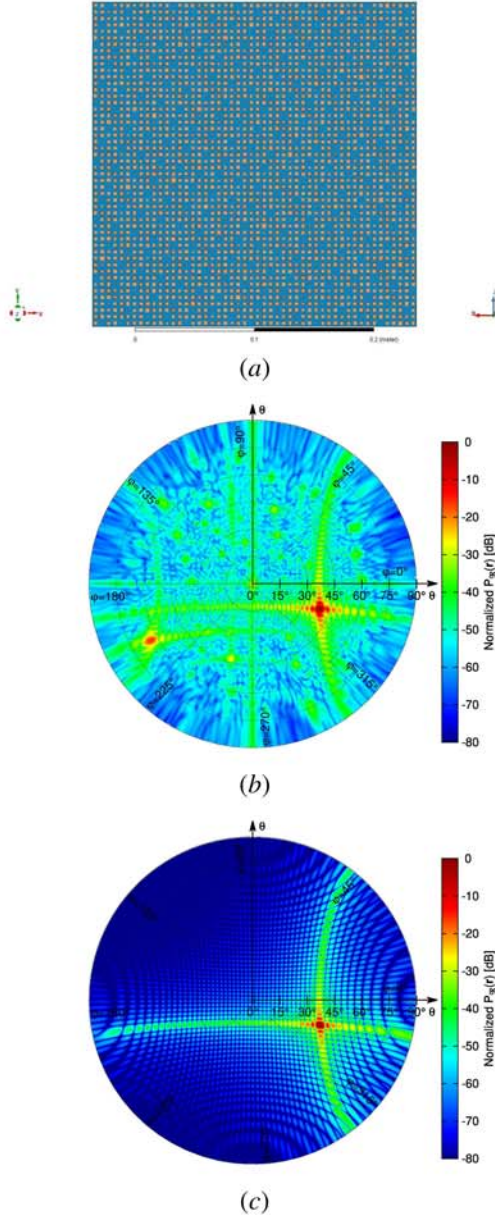


Figure 3. Numerical Validation. Layout of (a) the ANSYS HFSS model of the single-tile EMS and plot of (b)(c) the angular distribution of the power, $\mathcal{P}_R(\mathbf{r})$, reflected from the EMS on a sphere at a distance of 5 [m] and computed with (b) ANSYS HFSS or using (c) the closed-form relationship (2).

ones to be installed in \mathcal{S} , the problem is formulated as an optimization one by defining suitable optimization objectives aimed at quantifying the mismatch between the desired coverage (3) with that afforded by a trial EMS arrangement, \mathbf{T} , as well as the complexity of the final layout of the EMS, \mathbf{T}_{opt} . More specifically, the following two cost functions are defined. The former is the “coverage term”, $\Phi_1(\mathbf{T})$, given by

$$\Phi_1(\mathbf{T}) \triangleq \frac{1}{U} \sum_{u=1}^U \frac{\left| \sum_{n=1}^N t_n \mathcal{P}_R^{(n)}(\mathbf{r}_u; \mathbf{T}) - \mathcal{P}_{th} \right|}{\mathcal{P}_{th}} \times \mathcal{H} \left\{ \mathcal{P}_{th} - \sum_{n=1}^N t_n \mathcal{P}_R^{(n)}(\mathbf{r}_u; \mathbf{T}) \right\}, \quad (4)$$

while the latter is the “complexity term” defined as

$$\Phi_2(\mathbf{T}) \triangleq \frac{M}{N} \quad (5)$$

where U is the number of receivers, which are uniformly distributed inside the *AoI* Ω , and M ($M = \sum_{n=1}^N t_n$) is the number of tiles composing the EMS. Moreover, $\mathcal{H}(\cdot)$ is the Heaviside function equal to 1 when the argument is positive (i.e., $\mathcal{P}_{th} > \mathcal{P}_R(\mathbf{r}_u; \mathbf{T}) \rightarrow$ the power strength at \mathbf{r}_u is below the desired value) and 0 otherwise [i.e., $\mathcal{P}_{th} \leq \mathcal{P}_R(\mathbf{r}_u; \mathbf{T}) \rightarrow$ the coverage condition (3) is fulfilled].

Since the two cost functions to be minimized impose, on the one hand, the fitting of the coverage condition (4), while, on the other hand, the reduction of the number of tiles to minimize the area/cost of the skin (5), they are by definition conflicting. Indeed, a larger number of tiles leads to a stronger electric field in Ω and vice versa. Thus, the optimization problem turns out to be natively multi-objective and a natural solution strategy is that of defining a Pareto front of multiple optimal solutions, each being a valid trade-off to be considered for the final implementation of the EMS on the building facade. Such a multiplicity of solutions gives to the designer the possibility of choosing the EMS layout to be implemented according to its feeling and other non-functional constraints (e.g., architectural and landscaping restrictions, costs, etc ...). Following such a guideline, the binary NSGA-II [24] is chosen as optimization algorithm because of the binary nature of the design problem at hand (\mathbf{T} being a binary vector) and the need of synthesizing multiple trade-off solutions among the conflicting objectives. Moreover, thanks to its hill-climbing features [28], the Genetic Algorithm (GA) has global optimization features, which are here compulsory due to the non-convex behaviors of the cost functions. Indeed, Φ_1 and Φ_2 are non-continuous functions also characterized by the presence of local minima (i.e., sub-optimal solutions of the corresponding EMS design problem). More in detail, the following implementation of the binary NSGA-II is taken into account:

- **Step 0 - NSGA-II Setup.** Select the number of P individuals (i.e., trial layouts of the EMS) of the GA population and set the control parameters of the NSGA-II, namely the crossover rate, φ_c , the polynomial mutation rate, φ_m , the distribution index for both the crossover, κ_c , and the mutation rate, κ_m , and the maximum number of iterations, I , i being the iteration index ($i = 0, \dots, I$);
- **Step 1 - Population Initialization** ($i = 0$). Randomly set the initial trial solutions, $\{\mathbf{T}_i^{(p)}; p = 1, \dots, P\}$, and compute the cost function terms, $\Phi_{1,i}^{(p)} = \Phi_1(\mathbf{T}_i^{(p)})$ and $\Phi_{2,i}^{(p)} = \Phi_2(\mathbf{T}_i^{(p)})$, for each individual of the population ($p = 1, \dots, P$);
- **Step 2 - EMS Optimization** ($i = 1, \dots, I$). Apply the evolutionary operators of the NSGA-II to iteratively ($i \leftarrow i + 1$) generate the offsprings, $\{\mathbf{T}_i^{(p)}; p = 1, \dots, P\}$, from the current population of parents, $\{\mathbf{T}_{i-1}^{(p)}; p = 1, \dots, P\}$, and compute their fitness values (4) and (5). Stop the iterative process when the maximum number of iterations ($i = I$) is reached;

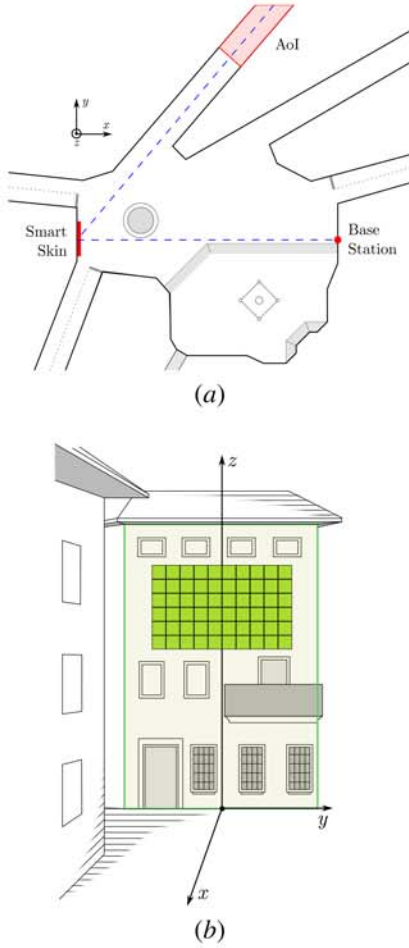


Figure 4. Numerical Validation - Simple Skin Layout - Orthogonal Incidence ($f = 27$ [GHz], $S = 15$ [m²], $L = 0.5$ [m], $N = 60$, $\Delta\Omega = 500$ [m²], $P_{th} = -70$ [dB]). Sketch of (a) the scenario and of (b) the admissible surface S along with its tile discretization.

- **Step 3 - Final Trade-Off EMS Design.** Select the set of O trial solutions that are non-dominated and belonging to the optimized Pareto front [24], $\{\mathbf{T}_{opt}^{(o)}; o = 1, \dots, O\}$. Such solutions are ordered according to the Φ_2 cost function value, namely $\mathbf{T}_{opt}^{(o)}|_{o=1} = \min_{o=1, \dots, O} \{\Phi_2(\mathbf{T}_{opt}^{(o)})\}$ and $\mathbf{T}_{opt}^{(o)}|_{o=O} = \max_{o=1, \dots, O} \{\Phi_2(\mathbf{T}_{opt}^{(o)})\}$. For a given o -th ($o = 1, \dots, O$) Pareto optimal solution, $\mathbf{T}_{opt}^{(o)}$, place the n -th ($n = 1, \dots, N$) tile $\mathcal{S}^{(n)}$ of the skin at the position $\mathbf{r}_S^{(n)}$ on the wall of the building if $t_{n,opt}^{(o)} = 1$.

IV. NUMERICAL RESULTS

The objective of this section is twofold. On the one hand, the critical evaluation of the impact on the wireless coverage of using modular EMSs. On the other, the assessment of the effectiveness of the proposed design method by considering different scenarios and tiles, while validating (2) through full-wave simulations with ANSYS HFSS [30].

As for this latter, the validation benchmark consists of a single ($N = 1$) tile located on the $y - z$ plane at the center of a Cartesian coordinate system and illuminated by a millimeter

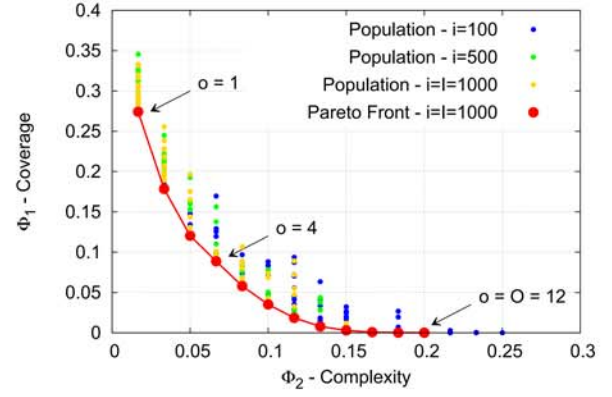


Figure 5. Numerical Validation - Simple Skin Layout - Orthogonal Incidence ($f = 27$ [GHz], $S = 15$ [m²], $L = 0.5$ [m], $N = 60$, $\Delta\Omega = 500$ [m²], $P_{th} = -70$ [dB]). Iterative ($i = 100$, $i = 500$, and $i = I$) evolution of the population of the P ($P = 2 \times N$) trial solutions, $\{\mathbf{T}_i^{(p)}; p = 1, \dots, P\}$, in the space of the design objectives and Pareto front at convergence ($i = I$), $\{\mathbf{T}_{opt}^{(o)}; o = 1, \dots, O\}$.

($f = 27$ [GHz]) plane wave generated from the BS (1) that impinges with an orthogonal incidence $(\theta_{\mathcal{S}}^{(1)}, \phi_{\mathcal{S}}^{(1)}) = (0, 0)$ [deg] on the tile with a vertical polarized (i.e., $\hat{\mathbf{e}}_r = \hat{\mathbf{y}}$) electric field. The distance between the BS and the tile has been chosen equal to $d_{\mathcal{S}}^{(1)} = 100$ [m] ($\rightarrow 9 \times 10^3 \lambda$), while the side and the square area of the tile has been set to $L^{(1)} = 25 \lambda$ and $\Delta S^{(1)} \simeq 0.277 \times 0.277$ [m²], respectively.

The realistic EMS tile of this specific example has been modeled with the ANSYS HFSS software as a metasurface defined by a lattice of unit cells uniformly-space along the x - and the y -axis by $\frac{\lambda}{2}$. Each unit-cell is composed by a square metallic patch printed on a single-layer Rogers 3003 dielectric substrate with thickness 5.08×10^{-4} [m] [Fig. 3(a)]. The single-tile EMS has been then designed by shaping the metallic patches of the metasurface² through the task-oriented multiscale design strategy based on the system-by-design paradigm proposed in [27] so that it reflects the impinging wave towards the direction $(\theta_{\mathcal{R}}^{(1)}, \phi_{\mathcal{R}}^{(1)}) = (40, -20)$ [deg].

For comparison purposes, the power, $\mathcal{P}_{\mathcal{R}}(\mathbf{r})$, reflected from the realistic [Fig. 3(b)] and the ideal [Fig. 3(c)] tiles on a sphere at a distance of 5 [m] from the skin barycenter $\mathbf{r}_S^{(0)}$ is shown in Fig. 3. As for the reflected electric field computed in HFSS, only the co-polar component is shown for the sake of comparison with the field reflected from ideal skin, which is not affected by the polarization loss [31]. Albeit the presence of undesired sidelobes, which are generally unavoidable when dealing with a real implementation of a low-cost artificial metasurface due to the differences between the ideal phase distribution of the theoretical model (2) and the actual one implemented in HFSS [Fig. 3(a)], the main lobes have a similar shape in both cases and they are steered towards the same desired angular direction [Figs. 3(b)-3(c)]. Since the pattern generated from the ideal tile according to (2) is

²It is worth to point out that each tile has to be different and characterized by a suitable layout of the metallic patches to enforce the required phase distribution in order to reflect the impinging wave along the desired direction.

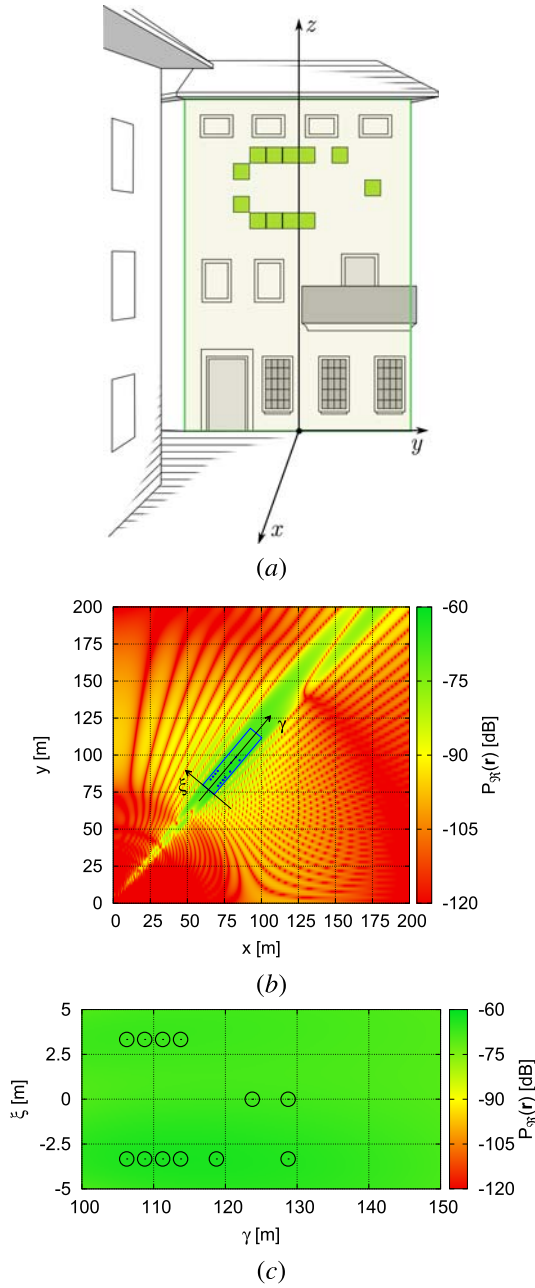


Figure 6. *Numerical Validation - Simple Skin Layout - Orthogonal Incidence* ($f = 27$ [GHz], $\mathcal{S} = 15$ [m²], $L = 0.5$ [m], $N = 60$, $\Delta\Omega = 500$ [m²], $\mathcal{P}_{th} = -70$ [dB]). Plot of (a) the EMS layout of the O -th ($O = 12$) solution, $\mathcal{S}_{opt}^{(O)}$, of the Pareto front in Fig. 5 and map of the spatial distribution of the power, $\mathcal{P}_R(r)$, reflected from the EMS in (b) the region Ψ and within (c) the AoI Ω ($\Omega \subset \Psi$).

free of undesired sidelobes and polarization losses, it seems reasonable to infer that the modular EMSs synthesized in this work will represent reference ideal solutions. From an operative viewpoint, this means that the number M of tiles composing the synthesized EMS layout has to be considered as a lower bound (i.e., the minimum number of tiles for approximating a project target) to be probably increased when going to the implementation of the EMS in a real scenario. Moving to the design of modular reflecting EMSs for an enhanced wireless coverage, the first test case (*Simple Skin*

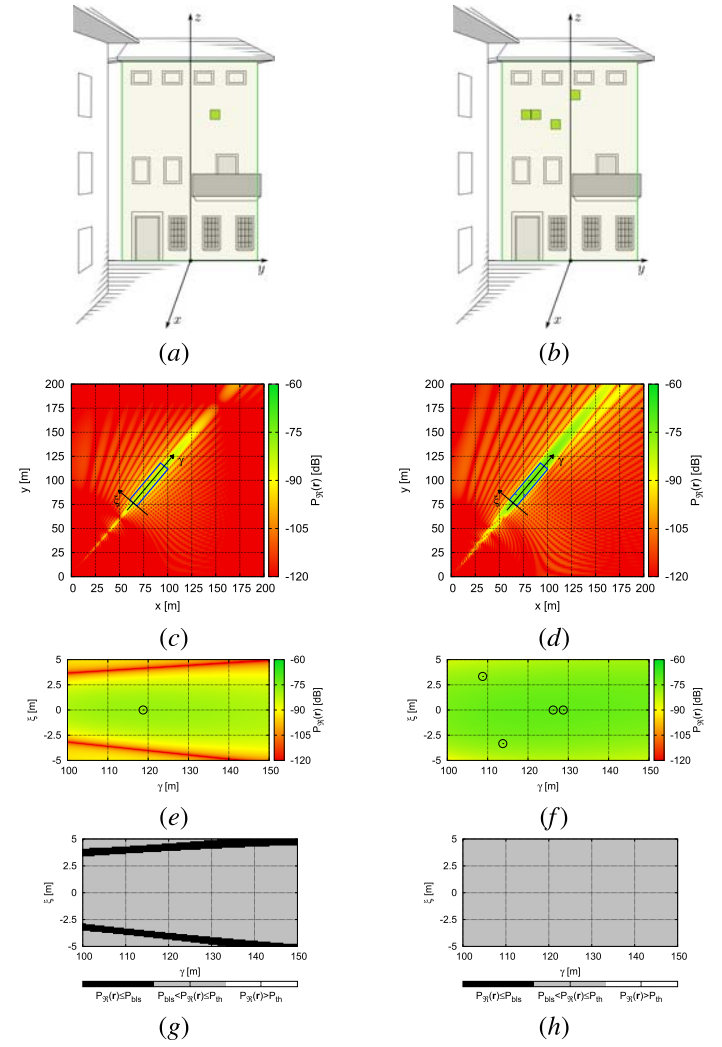


Figure 7. *Numerical Validation - Simple Skin Layout - Orthogonal Incidence* ($f = 27$ [GHz], $\mathcal{S} = 15$ [m²], $L = 0.5$ [m], $N = 60$, $\Delta\Omega = 500$ [m²], $\mathcal{P}_{th} = -70$ [dB], $\mathcal{P}_{bls} \approx -100$ [dB]). Plot of (a)(b) the EMS layout and of (c)(f) the corresponding spatial distributions of the power, $\mathcal{P}_R(r)$, reflected from the EMS along with (g)(h) the coverage/connectivity maps for (a)(c)(e)(g) the $o = 1$ and (b)(d)(f)(h) the $o = 4$ solutions of the Pareto front in Fig. 5.

Layout - Orthogonal Incidence) refers to the urban scenario depicted in Fig. 4(a) where the BS is located at \mathbf{r}_{BS} with Cartesian coordinates $(x_{BS}, y_{BS}, z_{BS}) = (100, 0, 10)$ [m]. The height of the BS from the ground (i.e., $z_{BS} = 10$ [m]) has been set according to the 3GPP guidelines for the *Urban-Micro (UMi)* cell scenario [32]. Moreover, the BS has been assumed to radiate a plane wave at $f = 27$ [GHz] having the electric field vertically-polarized ($\hat{\mathbf{e}}_r = \hat{\mathbf{z}}$) with unitary amplitude $E_{\mathcal{S}} = 1.0$ [V/m] that impinges from the ϕ -normal direction ($\phi_{\mathcal{S}}^{(0)} = 0.0$ [deg]) on the EMS placed on the $y - z$ plane [i.e., $x_{\mathcal{S}}^{(n)} = 0$ ($n = 1, \dots, N$)] [Fig. 4(b)] at a distance of $d_{\mathcal{S}}^{(0)} = 100$ [m] from the BS. The admissible surface \mathcal{S} for the deployment of the artificial EMS on the building facade [Fig. 4(b)] has been chosen with an area $\Delta\mathcal{S} = 15$ [m²] and it extends within the range -2.5 [m] $\leq y_{\mathcal{S}} \leq 2.5$ [m] and 5.0 [m] $\leq z_{\mathcal{S}} \leq 8.0$ [m] along the y - and the z -axis, respectively. Such an area \mathcal{S} has been partitioned into $N = 60$ square

Table I
Numerical Validation. STATISTICS OF THE REFLECTED POWER $\mathcal{P}_{\mathcal{R}}(\mathbf{r})$ WITHIN THE AoI, Ω .

	$\min_{\mathbf{r} \in \Omega} \{ E_{rx}(\mathbf{r}) ^2\}$ [dB]	$\max_{\mathbf{r} \in \Omega} \{ E_{rx}(\mathbf{r}) ^2\}$ [dB]	$\text{avg}_{\mathbf{r} \in \Omega} \{ E_{rx}(\mathbf{r}) ^2\} = \Phi_1$ [dB]
Fig. 6(c)	-69.9	-63.0	-66.8
Fig. 7(e)	-173.9	-78.1	-84.7
Fig. 7(f)	-90.9	-71.7	-75.0
Fig. 9(c)	-69.9	-63.2	-66.9
Fig. 12(g)	-69.7	-62.5	-65.1
Fig. 12(h)	-69.7	-61.2	-64.4
Fig. 12(i)	-101.5	-60.3	-65.1
Fig. 14(c)	-73.1	-60.1	-63.5

sub-domains of size $\Delta\mathcal{S}^{(n)} = L^{(n)} \times L^{(n)}$ ($n = 1, \dots, N$) being $L^{(n)} = 0.5$ [m], so that there are $N_y = 10$ and $N_z = 6$ partitions along the y - and the z -axis, respectively (i.e., $N = N_y \times N_z$). By enumerating the admissible locations of the EMS tiles in a raster scan way, starting from the top left corner of \mathcal{S} , the barycenter of the n -th ($n = 1, \dots, N$) sub-domain of \mathcal{S} has the following coordinates

$$\begin{aligned} y_S^{(n)} &= y_S^{(1)} + \left(n - 1 - \left\lfloor \frac{n-1}{N_y} \right\rfloor N_y\right) \times L^{(n)} \\ z_S^{(n)} &= z_S^{(1)} - \left\lfloor \frac{n-1}{N_y} \right\rfloor \times L^{(n)} \end{aligned} \quad (6)$$

($n = 2, \dots, N$), $y_S^{(1)} = -2.25$ [m] and $z_S^{(1)} = 7.75$ [m] being the coordinates of the barycenter of the first ($n = 1$) location admissible for a tile.

The goal of the EMS design is that of enhancing the power strength in the AoI Ω of size $\Delta\Omega = 10 \times 50$ [m²] located in $x_\Omega^{(0)} = 80.35$ [m] and $y_\Omega^{(0)} = 95.75$ [m] [Fig. 4(a)] along the azimuth direction $\phi_\Omega^{(0)} = 50.0$ [deg] with respect to the EMS, $\mathcal{P}_{th} = -70$ [dB] being the threshold on the desired coverage in (4), while the “no connection” power level has been set to $\mathcal{P}_{bls} \approx -100$ [dB]. In order to assess the “coverage” condition within the AoI, $U = 500$ ideal receivers have been uniformly-distributed within Ω , that is one receiver every $\Delta\Omega = 1$ [m²], at the height $z_u = 1.5$ [m] to emulate users on the ground [32].

Concerning the NSGA-II algorithm, the following setup of the control parameters has been used: $P = 2 \times N$, $I = 1000$, $\wp_c = 1.0$, $\wp_m = \frac{1}{N}$, $\aleph_c = 15$, and $\aleph_m = 20$. Moreover, each simulation has been repeated 50 times with different random seeds to statistically validate the results from the stochastic optimization. However, since all simulations led to similar Pareto fronts at convergence ($i = I$), only the solutions of a representative run will be reported and discussed in the following.

Figure 5 shows the population of trial solutions, $\{\mathbf{T}_i^{(p)}; p = 1, \dots, P\}$, at the iterations $i = 100$, $i = 500$, and $i = I$ in the space of the objectives along with the Pareto front of $O = 12$ non-dominated solutions at the convergence ($i = I$), $\{\mathbf{T}_{opt}^{(o)}; o = 1, \dots, O\}$. Let us now analyze the EMS solution that fully fits the coverage requirements [i.e., $\Phi_1(\mathbf{T}_{opt}^{(o)}) = 0$], which corresponds to the O -th representative point of the Pareto front, whose chromosome has $M = 12$

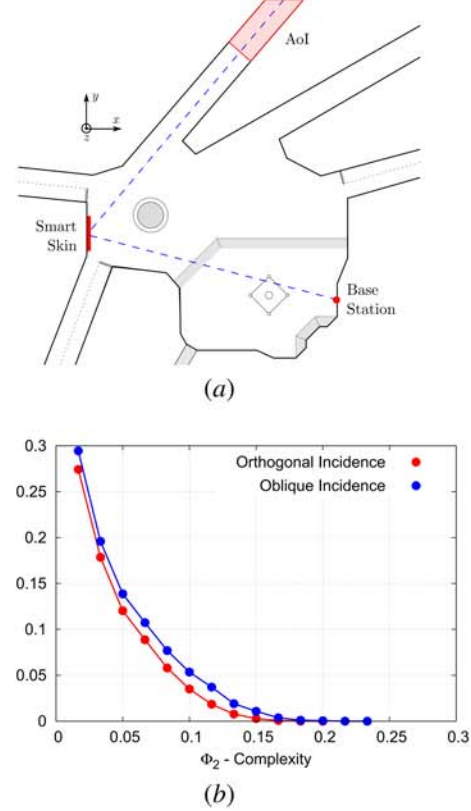


Figure 8. Numerical Validation - Simple Skin Layout - Oblique Incidence ($f = 27$ [GHz], $\mathcal{S} = 15$ [m²], $L = 0.5$ [m], $N = 60$, $\Delta\Omega = 500$ [m²], $\mathcal{P}_{th} = -70$ [dB]). Sketch of (a) the scenario and plot of (b) the Pareto front at convergence ($i = I$), $\{\mathbf{T}_{opt}^{(o)}; o = 1, \dots, O\}$.

bits at one (i.e., $t_{n,opt}^{(o)} \Big|_{o=12} = 1$, $n = \{3, 4, 5, 6, 8, 12, 30, 32, 43, 44, 45, 46\}$) so that the EMS layout turns out being composed by $M = 12$ tiles as shown in Fig. 6(a). The support of such an EMS amounts to $\Delta\mathcal{S}_{opt}^{(o)} \Big|_{o=12} = 3$ [m²], which is one fifth of the whole admissible EMS surface (i.e., $\frac{\Delta\mathcal{S}_{opt}^{(o)} \Big|_{o=12}}{\Delta\mathcal{S}} = 20\%$). The coverage improvement enabled by the installation of such an artificial skin on the building facade is pointed out in Fig. 6(b) where the map of the reflected power $\mathcal{P}_{\mathcal{R}}(\mathbf{r})$ at $z = 1.5$ [m] from the ground is shown in a region Ψ , around the AoI, of extension $\Delta\Psi = 200 \times 200$ [m²]. As it can be observed, the power intensity along the direction of the street passing through

Ω has been significantly increased. As a matter of fact, the signal turns out to be stronger not only in Ω , but also before and after [Fig. 6(b)] since the skin tiles generate simple pencil beams (2) with an elongated footprint on the ground, the dots in Figs. 6(b)-6(c) being the $M = 12$ points $\mathbf{r}_\Omega^{(n)}$, $n = \{3, 4, 5, 6, 8, 12, 30, 32, 43, 44, 45, 46\}$ where the peaks of the beams reflected from the M EMS tiles are directed. Regardless of the simplicity of the beam afforded by a single EMS tile, the combined use of multiple/modular tiles has allowed to reach the desired average power threshold $\mathcal{P}_{th} = -70$ [dB] (i.e., $\Phi_1(\mathbf{T}_{opt}^{(o)})|_{o=12} = 0$) in the whole AoI as shown in Fig. 6(c), $\gamma - \xi$ being the Ω local coordinate system [see Fig. 6(b)]. Indeed, the statistics of the power reflected in Ω are: $\min_{\mathbf{r} \in \Omega} \{\mathcal{P}_R(\mathbf{r}_u; \mathbf{T}_{opt}^{(o)})|_{o=12}\} = -69.9$ [dB], $\max_{\mathbf{r} \in \Omega} \{\mathcal{P}_R(\mathbf{r}_u; \mathbf{T}_{opt}^{(o)})|_{o=12}\} = -63.0$ [dB], and $\text{avg}_{\mathbf{r} \in \Omega} \{\mathcal{P}_R(\mathbf{r}_u; \mathbf{T}_{opt}^{(o)})|_{o=12}\} = -66.8$ [dB], respectively (Tab. I).

For the sake of completeness, other two representative solutions of the Pareto front in Fig. 5 are analyzed. The EMS layouts and the maps of the power reflected in Ψ of the solution with minimum complexity ($o = 1$ - Fig. 5) and the one having $\Phi_1(\mathbf{T}_{opt}^{(o)})|_{o=4} \approx 0.1$ (i.e., $\text{avg}_{\mathbf{r} \in \Omega} \{\mathcal{P}_R(\mathbf{r}_u; \mathbf{T}_{opt}^{(o)})|_{o=12}\} = -77$ [dB]) are reported in Fig. 7.

The minimum complexity EMS (i.e., $\Phi_2(\mathbf{T}_{opt}^{(o)})|_{o=1} = \frac{1}{60}$) needs only one ($M = 1$) tile [Fig. 7(a)], but the average power level in Ω reduces of $\delta\Phi_1|_{o=1}^{o=12} = 17.9$ [dB] ($\delta\Phi|_{o'} \triangleq \Phi(\mathbf{T}_{opt}^{(o)}) - \Phi(\mathbf{T}_{opt}^{(o')})$; $o, o' \in [1, O]$) with respect to that in Fig. 6(a). Owing to the presence of a single tile, the map of the EM power in Fig. 7(c) shows the classical footprint of a pencil beam characterized by a mainlobe focused in the point $\mathbf{r}_\Omega^{(n)}|_{n=28}$ within Ω [Fig. 7(c)] along the central line of the AoI ($\xi = 0$) [Fig. 7(e)], while there are portions of Ω close to $\xi = \pm 5$ [m] where the power strength is very low [Fig. 7(e)]. Quantitatively, it turns out that the condition $\mathcal{P}_{th} \leq \mathcal{P}_R(\mathbf{r}_u; \mathbf{T}_{opt}^{(o)})|_{o=1}$ never holds true [Fig. 7(e)] since also the power peak is below the desired QoS threshold ($\mathcal{P}_R(\mathbf{r}_u; \mathbf{T}_{opt}^{(o)})|_{o=1} = -78.1$ [dB]). Moreover, the minimum level of the power reflected within the AoI by such a single-tile EMS of side $L^{(1)} = 0.5$ [m] is equal to $\min_{\mathbf{r} \in \Omega} \{\mathcal{P}_R(\mathbf{r}_u; \mathbf{T}_{opt}^{(o)})|_{o=1}\} = -173.9$ [dB] (Tab. I), that is (well) below the “connectivity” threshold of $\mathcal{P}_{bls} = -100$ [dB]. Such an undesired condition verifies in other portions of Ω where there is not enough signal for assuring the users’ connections [$\mathcal{P}_R(\mathbf{r}_u; \mathbf{T}_{opt}^{(o)})|_{o=1} < \mathcal{P}_{bls}$ - Fig. 7(g)].

By using three more tiles [i.e., $M = 4$ - Fig. 7(b)], the power level reflected on Ω turns out significantly higher [Fig. 7(d)], the average power being increased of almost ten times (i.e. $\delta\Phi_1|_{o=1}^{o=4} = 9.7$ [dB]), and there are no more “no connection” zones within Ω [Fig. 7(h)] since $\min_{\mathbf{r} \in \Omega} \{\mathcal{P}_R(\mathbf{r}_u; \mathbf{T}_{opt}^{(o)})|_{o=4}\} = -90.9$ [dB] (Tab. I), even though the power peak is still slightly lower than the QoS threshold

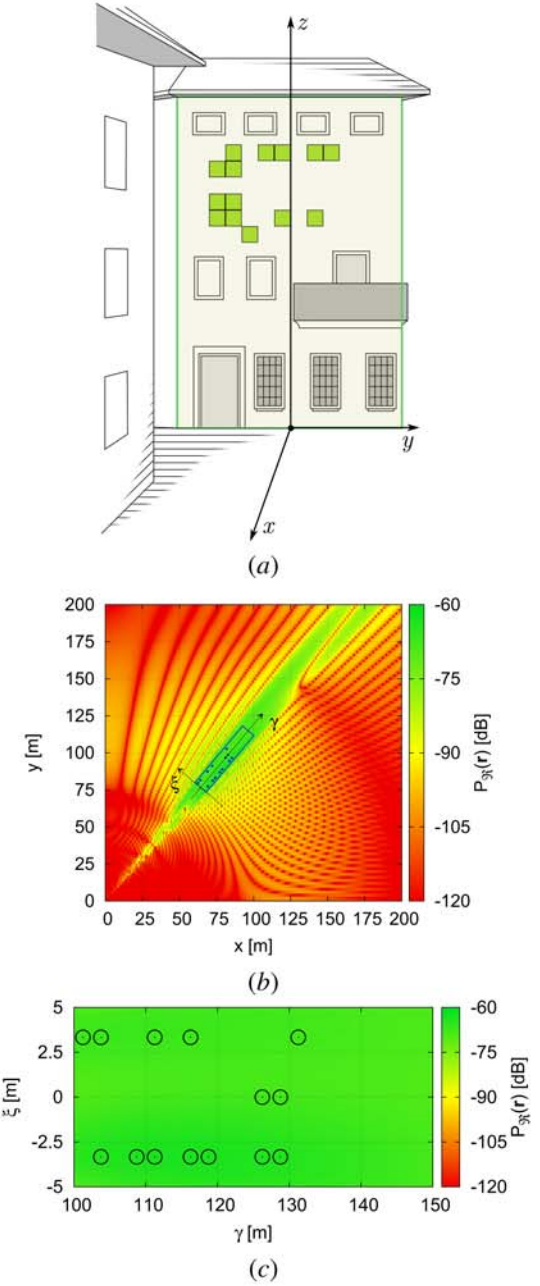


Figure 9. Numerical Validation - Simple Skin Layout - Oblique Incidence ($f = 27$ [GHz], $S = 15$ [m²], $L = 0.5$ [m], $N = 60$, $\Delta\Omega = 500$ [m²], $\mathcal{P}_{th} = -70$ [dB]). Plot of (a) the EMS layout of the O -th ($O = 14$) solution, $\mathcal{S}_{opt}^{(O)}$, of the Pareto front in Fig. 5 and map of the spatial distribution of the power, $\mathcal{P}_R(\mathbf{r})$, reflected from the EMS in (b) the region Ψ and within (c) the AoI Ω ($\Omega \subset \Psi$).

$$(\max_{\mathbf{r} \in \Omega} \{\mathcal{P}_R(\mathbf{r}_u; \mathbf{T}_{opt}^{(o)})|_{o=4}\}) = -71.7$$
 [dB].

In the second test case (Simple Skin Layout - Oblique Incidence), the field generated from the BS has been assumed impinging on the EMS with an oblique incidence on the azimuth plane (i.e., $\phi_\Omega^{(0)} = 20$ [deg]), being $(x_{BS}, y_{BS}, z_{BS}) = (93.9, 34.2, 10)$ [m] such that $d_\Omega^{(0)} = 100$ [m] as in the previous example [Fig. 8(a)]. All other features concerned with the EM field generated from the BS (i.e., frequency and polarization), the area and the discretization of \mathcal{S} , and the coverage area under analysis Ω have been kept unaltered from

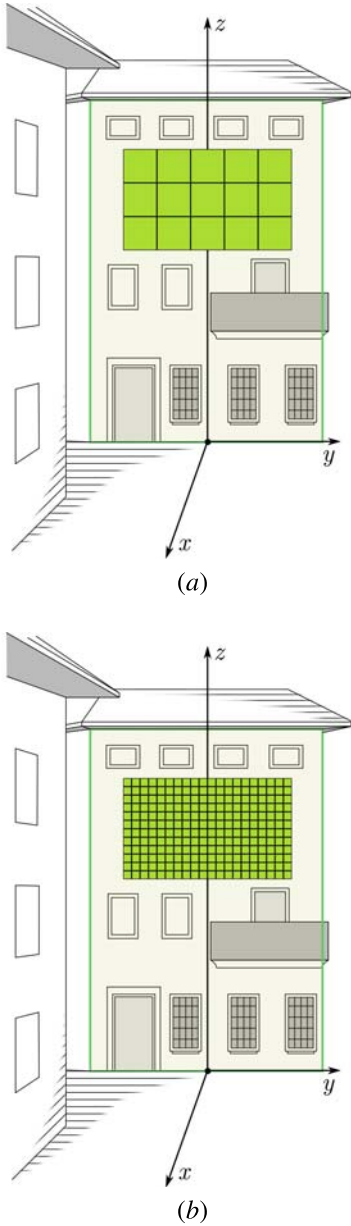


Figure 10. *Numerical Validation - Simple Skin Layout - Varying Tiles Size* ($f = 27$ [GHz], $S = 15$ [m²], $N = \{240, 15\}$, $\mathcal{P}_{th} = -70$ [dB]). Sketches of the scenario and of the admissible surface \mathcal{S} along with its discretization when using tiles with side-length (a) $L = 1.0$ [m] and (b) $L = 0.25$ [m].

the first test case.

The Pareto front of the O optimal trade-off solutions determined by the proposed *NSGA-II* based approach is compared in Fig. 8(b) with that obtained in the “normal incidence” case (Fig. 5). The reader can observe that the Pareto front of the “oblique incidence” scenario consists of $O = 14$ *EMS* designs (vs. $O = 12$ - “normal incidence”) and it turns out that the oblique incidence from the *BS* needs a higher number of tiles to yield the same coverage of the “normal incidence” solutions [Fig. 8(b)]. As expected, a wider area is now required because of the reduction of the effective area of the *EMS* (2) since a lower amount of power is intercepted from the incident wave with the same area of the “normal incidence” *EMS*.

Figure 9 summarizes the characteristics of the O -th

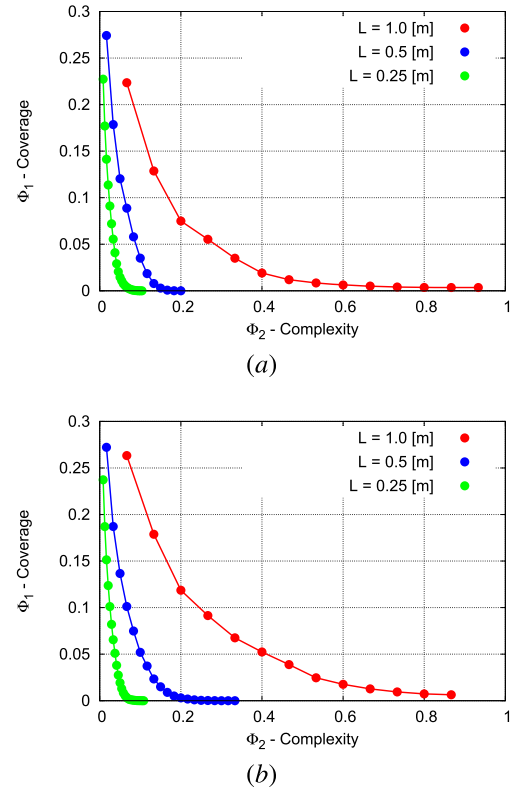


Figure 11. *Numerical Validation - Simple Skin Layout - Varying Tiles Size* ($f = 27$ [GHz], $S = 15$ [m²], $\mathcal{P}_{th} = -70$ [dB]). Plot of the Pareto fronts at convergence ($i = I$), $\{\mathbf{T}_{opt}^{(o)}; o = 1, \dots, O\}$, for different square tile sizes (L being the side-length of the tile) in correspondence with an *AoI* Ω of dimension (a) $\Delta\Omega = 10 \times 50 = 500$ [m²] and (b) $\Delta\Omega = 10 \times 100 = 1000$ [m²].

($O = 14$) solution that fits the coverage requirement [i.e., $\mathcal{P}_{\mathcal{R}}(\mathbf{r}_u; \mathbf{T}_{opt}^{(O)}) \geq \mathcal{P}_{th} \rightarrow \Phi_1(\mathbf{T}_{opt}^{(O)}) = 0$]. More in detail, the layout of the corresponding *EMS* is composed by $M = 14$ tiles [Fig. 9(a)], that is two more than those of the O -th *EMS* for the normal incidence, and the arising tiles arrangement is also quite different [Fig. 9(a) vs. Fig. 6(a)]. On the contrary, the power distributions are quite similar as pictorially shown by comparison of the maps in Figs. 9(b)-9(c) with those in Figs. 6(b)-6(c) and also confirmed by the statistics of the power reflected by the *EMS* within the *AoI* (Tab. I). Indeed, the differences among the values of the statistical indices are null or negligible (i.e., $\delta\mathcal{P}_{\mathcal{R}}^{min}{}_{Normal}^{Oblique} = 0.0$ [dB], $\delta\mathcal{P}_{\mathcal{R}}^{max}{}_{Normal}^{Oblique} = 0.2$ [dB], and $\delta\mathcal{P}_{\mathcal{R}}^{av}{}_{Normal}^{Oblique} = \delta\Phi_1{}_{Normal}^{Oblique} = -0.1$ [dB], being $\delta\mathcal{P}_{\mathcal{R}}^{stat}{}_{Normal}^{Oblique} \triangleq \text{stat}_{\mathbf{r} \in \Omega} \left\{ \mathcal{P}_{\mathcal{R}}(\mathbf{r}_u; \mathbf{T}_{opt}^{(o)})_{o=O} \right\} - \text{stat}_{\mathbf{r} \in \Omega} \left\{ \mathcal{P}_{\mathcal{R}}(\mathbf{r}_u; \mathbf{T}_{opt}^{(o)})_{o=O}^{Normal} \right\}$ - Tab. I).

The third design experiment (*Simple Skin Layout - Varying Tiles Size*) is concerned with a surface \mathcal{S} still discretized with uniform tiles, but considering different tile sizes: $L^{(n)} = 1.0$ [m] [$\rightarrow N = N_y \times N_z = 5 \times 3 = 15$ - Fig. 10(a)] or $L^{(n)} = 0.25$ [m] [$\rightarrow N = N_y \times N_z = 20 \times 12 = 240$ - Fig. 10(b)] ($n = 1, \dots, N$). In the former case, there are few admissible tiles reflecting a narrow beam towards the *AoI*, while the number of tiles and *DoFs* is 16 times larger in the latter case where

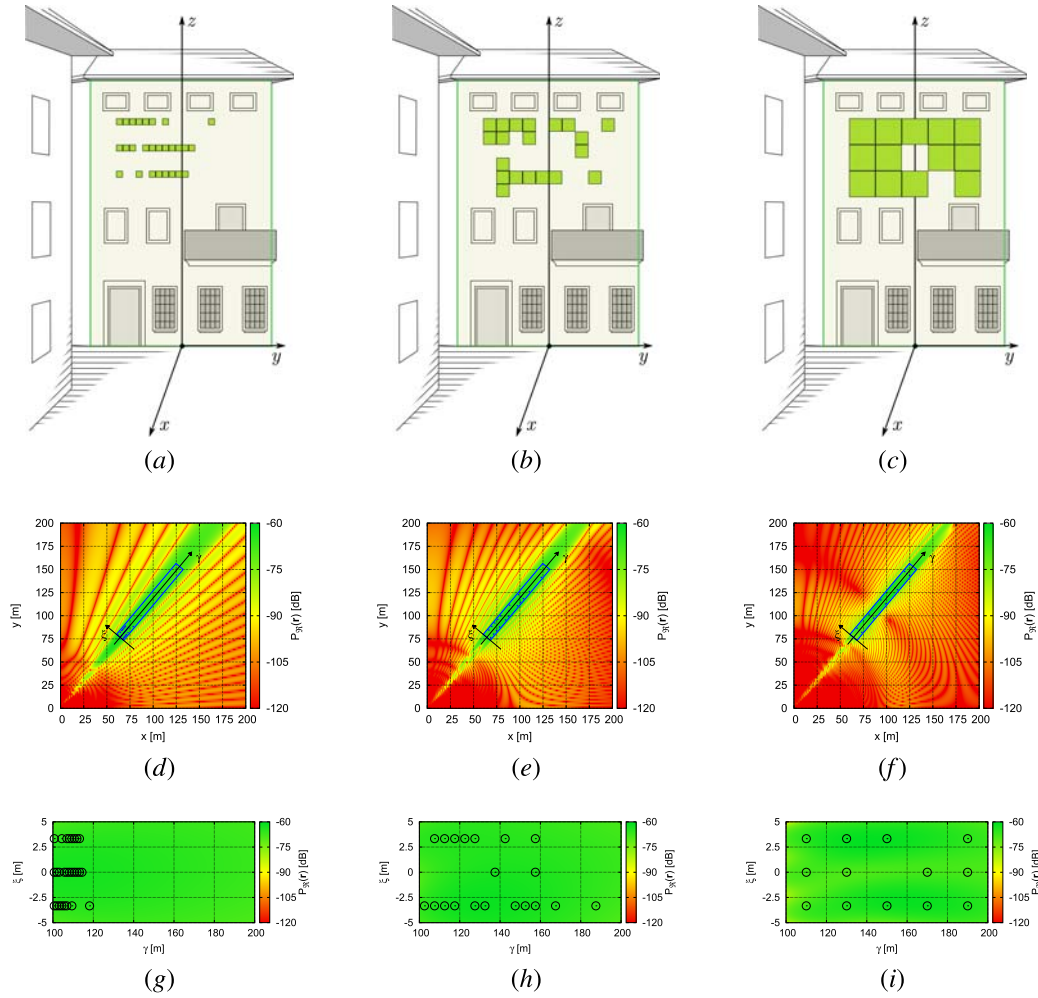


Figure 12. Numerical Validation - Simple Skin Layout - Varying Tiles Size ($f = 27$ [GHz], $S = 15$ [m²], $L = \{0.25, 0.5, 1.0\}$ [m], $N = \{240, 60, 15\}$, $\Delta\Omega = 1000$ [m²], $\mathcal{P}_{th} = -70$ [dB]). Plot of (a)-(c) the EMS layouts of the O -th solution, $\mathcal{S}_{opt}^{(O)}$, of the Pareto fronts in Fig. 11(b) and maps of the spatial distribution of the power, $\mathcal{P}_R(\mathbf{r})$, reflected from the EMS in (d)-(f) the region Ψ and within (g)-(i) the AoI Ω ($\Omega \subset \Psi$) when using tiles with side-length (a)(d)(g) $L = 0.25$ [m], (b)(e)(h) $L = 0.5$ [m], and (c)(f)(i) $L = 1.0$ [m].

the beam reflected by each m -th ($m = 1, \dots, M$) installed tile has a broader coverage.

The NSGA-II optimization has been run for both tile sizes and the Pareto fronts obtained at the convergence ($i = I$) are shown in Fig. 11(a) along with that of Fig. 5, which is related to the tile size $L^{(n)} = 0.5$ [m] ($n = 1, \dots, N$). The plots in Fig. 11(a) indicate that, the wider the tile size, the higher is the value of the complexity index Φ_2 to fulfil (3). Furthermore, it is worth highlighting that when $L^{(n)} = 1.0$ [m], no solution of the Pareto front satisfies the coverage requirement since $\Phi_1(\mathbf{T}_{opt}^{(O)}) > 0$, while the coverage condition holds true using smaller tiles.

The same conclusions arise when extending the coverage area Ω from $\Delta\Omega = 10 \times 50$ [m²] [Fig. 11(a)] up to $\Delta\Omega = 10 \times 100$ [m²] [Fig. 11(b)]. For this latter case, the layouts and the coverage maps of the solutions providing the best coverage (i.e., the O -th of the Pareto front) for each tile size are shown in Fig. 12. By analyzing the power distributions in Figs. 12(d)-12(f), it turns out that the main advantage of using larger tiles, which reflect narrower beams, is the capability of better

focusing the reflected field only along the direction of the AoI [Fig. 12(f) vs. Fig. 12(d)]. This is not for free and the cost to pay is that of having a very large EMS composed by $M = 13$ tiles, each of $\Delta\mathcal{S}^{(m)} = 1$ [m²] ($m = 1, \dots, M$), for a total surface of $\Delta\mathcal{S}_{opt}^{(O)} \Big|_{L=1.0 \text{ [m]}} = 13$ [m²], while the area occupied by the EMS when using square tiles of size $L^{(n)} = 0.25$ [m] [Fig. 12(a)] and $L^{(n)} = 0.5$ [m] [Fig. 12(b)] amounts to $\Delta\mathcal{S}_{opt}^{(O)} \Big|_{L=0.25 \text{ [m]}} \simeq 1.69$ [m²] and $\Delta\mathcal{S}_{opt}^{(O)} \Big|_{L=0.5 \text{ [m]}} = 5$ [m²], respectively. This means that the limited focusing capability of smaller tiles is balanced by a reduction of the required EMS extension, thus a lower cost of the EMS.

In the last design example (*Complex Skin Layout - Orthogonal Incidence*), the admissible region \mathcal{S} on the building facade is more complex [Fig. 13(a)] since the area dedicated to the EMS deployment is smaller and there are more architectural constraints (e.g., misaligned windows and open window shutters) as in historical buildings. As for the descriptive parameters of the scenario at hand, they have been defined as in the “Simple

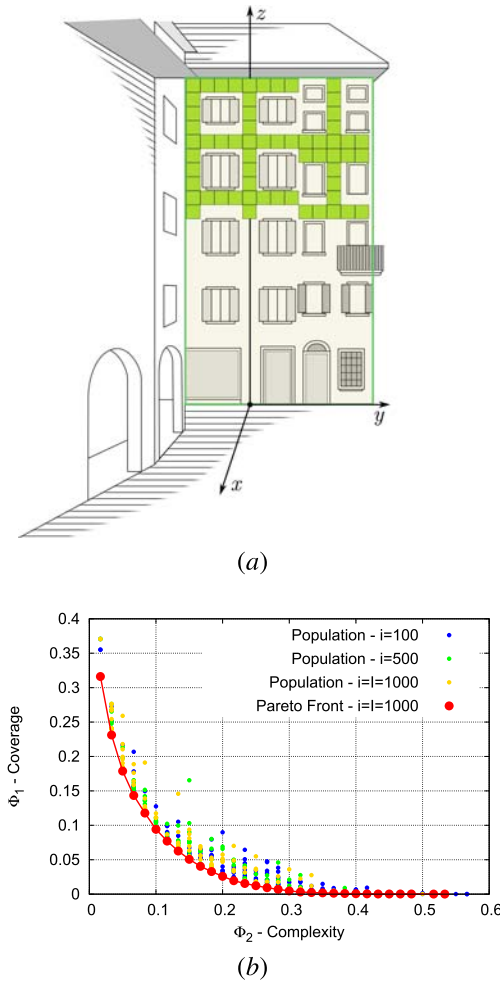


Figure 13. Numerical Validation - Complex Skin Layout - Orthogonal Incidence ($f = 27$ [GHz], $S = 15$ [m²], $L = 0.5$ [m], $N = 60$, $\Delta\Omega = 1000$ [m²], $\mathcal{P}_{th} = -70$ [dB]). Sketch of (a) the scenario and of the admissible surface S along with its tile discretization and plot of (b) the iterative ($i = 100$, $i = 500$, and $i = I$) evolution of the population of the P ($P = 2 \times N$) trial solutions, $\{\mathbf{T}_i^{(p)}; p = 1, \dots, P\}$, in the space of the design objectives along with the Pareto front at convergence ($i = I$), $\{\mathbf{T}_{opt}^{(o)}; o = 1, \dots, O\}$.

Skin Layout - Orthogonal Incidence” case, but a larger *AoI* (i.e., $\Delta\Omega = 10 \times 100$ [m²]) has been considered.

The evolution ($i = 100$, $i = 500$, and $i = I$) of the population of *EMS* trial solutions, $\{\mathbf{T}_i^{(p)}; p = 1, \dots, P\}$, in the space of the objectives is shown in Fig. 13(b) together with the Pareto front at convergence ($i = I$), which includes $O = 31$ non-dominated solutions, $\{\mathbf{T}_{opt}^{(o)}; o = 1, \dots, O\}$. As it can be noticed [Fig. 13(b)], 15 *EMS* of the Pareto front have values of the coverage index smaller than $\Phi_1(\mathbf{T}_{opt}^{(o)}) < 10^{-2}$.

The O -th solution, which fully satisfies the coverage requirements, is done by $M = 32$ tiles of size $\Delta S^{(m)} = 0.5 \times 0.5$ [m²] ($m = 1, \dots, M$) and it covers a surface area of $\Delta S_{opt}^{(O)} = 8$ [m²] [Fig. 14(a)]. Despite the irregularity of the *EMS* layout, the coverage maps in Figs. 14(b)-14(d) confirm that the proposed *EMS* design method properly selects, from the admissible pool, a subset of tiles that guarantees the required power level within the *AoI* Ω (Tab. I).

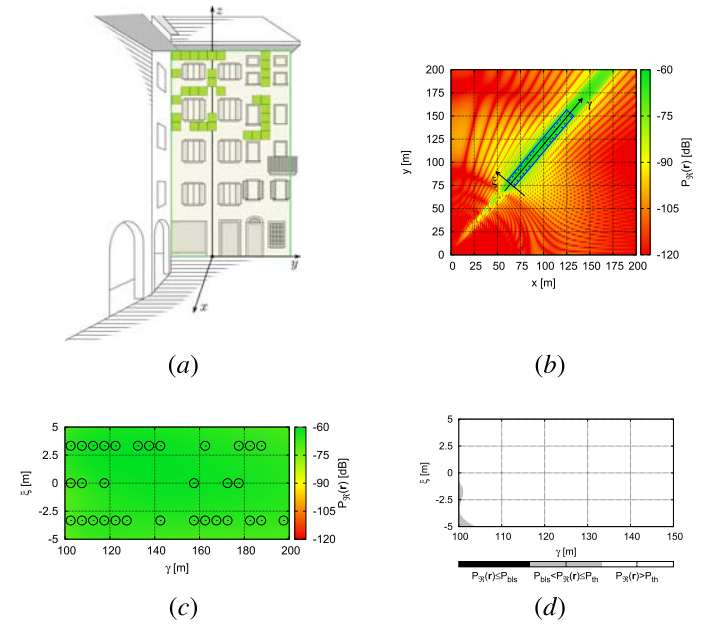


Figure 14. Numerical Validation - Complex Skin Layout - Orthogonal Incidence ($f = 27$ [GHz], $S = 15$ [m²], $L = 0.5$ [m], $N = 60$, $\Delta\Omega = 1000$ [m²], $\mathcal{P}_{th} = -70$ [dB]). Plot of (a) the *EMS* layout and of (b)(c) the corresponding spatial distributions of the power, $\mathcal{P}_R(\mathbf{r})$, reflected from the *EMS* along with (d) the coverage/connectivity maps for the O -th ($O = 31$) solution of the Pareto front in Fig. 13(b).

V. CONCLUSIONS

Within the *SEME* vision, this paper has proposed a novel strategy to improve the signal strength in urban areas where the power radiated by the *BS* is too strongly attenuated. More specifically, such an approach proposes the use of modular, passive, and static artificial metasurfaces to be installed/embedded on the facades of urban buildings, such as coating skins, to enhance the coverage by reflecting the *EM* wave coming from the *BS* towards the desired directions within an *AoI*. In order to fulfil user-defined coverage conditions, while minimizing the cost/complexity, the design of the *EMS* has been cast as a multi-objective optimization problem and it has been addressed by means of a binary implementation of the *NSGA-II* algorithm.

From a technological and methodological viewpoint, the main novelties, to the best of the authors’ knowledge, of this research work can be summarized as follows:

- the introduction for the first time of a novel cost-effective solution, to be possibly implemented through cheap printed technology, for the large scale deployment of artificial metasurfaces to be installed on the facades of buildings for improving the wireless coverage in urban scenarios;
- the suitability of the proposed technological solution in future wireless networks thanks to its “green” (i.e., passive) and non-invasive (i.e., low profile and without heavy architectural impact) nature;
- the development of a customized design strategy to enable an effective/efficient optimization-based design of (also large) *EMS*s composed by simple (also non-homogeneous) tiles.

From the numerical assessment, which has been carried out by considering realistic topological urban scenarios and a millimeter-wave 5G frequency band, the following outcomes can be drawn:

- the use of tiled *EMS*s always improves the coverage of the *AoI*;
- the (*NSGA-II*)-based synthesis approach provides the designer with a Pareto front of multiple *EMS* solutions, which are trade-offs between coverage requirements and complexity of the *EMS* layout. In all the considered scenarios, an *EMS* that fulfils the user-defined coverage condition (i.e., not only the user connection) has been (generally) found without using the whole area available on the facade of the building;
- the number and the positions of the tiles of the *EMS* layout depend on the relative position between the *BS* and the *AoI*. Moreover, the dimension (e.g., small, medium, large) and the distribution (e.g., uniform or non-uniform) of the tiles composing the *EMS* are other *DoFs*, which can be exploited to fit the coverage conditions as well as other architectural constraints (e.g., misaligned windows and open window shutters).

Future research activities, beyond the scope of this paper, will integrate the design of the tiles layout within the proposed iterative optimization loop in order to take into account non-ideal reflections and polarization losses, and the possibility to design more complex (e.g., conformal and not flat) skin surfaces. Moreover, the presence of multiple *BS*s and *AoI*s will be dealt with towards the definition of a tool for network planning and the use of such skins will be also investigated for improving the coverage indoor.

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